

# **ANEXO II**

## **(Hojas técnicas de los dispositivos)**

	Proyecto Fin de Carrera	Alumno
	Diseño e implementación de un convertidor monofásico de cinco niveles con control basado en DSP	José Francisco Campos Bizcocho

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## 1. IRG4PH40KD.

International  
**IR Rectifier**

PD- 91577B

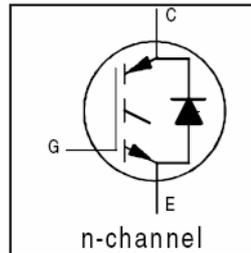
# IRG4PH40KD

INSULATED GATE BIPOLAR TRANSISTOR WITH  
ULTRAFAST SOFT RECOVERY DIODE

Short Circuit Rated  
UltraFast IGBT

### Features

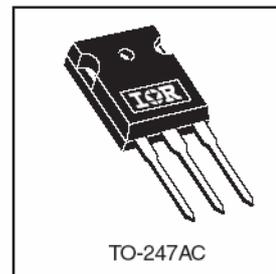
- High short circuit rating optimized for motor control,  $t_{sc} = 10\mu s$ ,  $V_{CC} = 720V$ ,  $T_J = 125^\circ C$ ,  $V_{GE} = 15V$
- Combines low conduction losses with high switching speed
- Tighter parameter distribution and higher efficiency than previous generations
- IGBT co-packaged with HEXFRED™ ultrafast, ultrasoft recovery antiparallel diodes



$V_{CES} = 1200V$
$V_{CE(on) typ.} = 2.74V$
@ $V_{GE} = 15V, I_C = 15A$

### Benefits

- Latest generation 4 IGBT's offer highest power density motor controls possible
- HEXFRED™ diodes optimized for performance with IGBTs. Minimized recovery characteristics reduce noise, EMI and switching losses
- This part replaces the IRGPH40KD2 and IRGPH40MD2 products
- For hints see design tip 97003



### Absolute Maximum Ratings

	Parameter	Max.	Units
$V_{CES}$	Collector-to-Emitter Voltage	1200	V
$I_C @ T_C = 25^\circ C$	Continuous Collector Current	30	A
$I_C @ T_C = 100^\circ C$	Continuous Collector Current	15	
$I_{CM}$	Pulsed Collector Current ①	60	
$I_{LM}$	Clamped Inductive Load Current ②	60	
$I_F @ T_C = 100^\circ C$	Diode Continuous Forward Current	8.0	
$I_{FM}$	Diode Maximum Forward Current	130	
$t_{sc}$	Short Circuit Withstand Time	10	$\mu s$
$V_{GE}$	Gate-to-Emitter Voltage	$\pm 20$	V
$P_D @ T_C = 25^\circ C$	Maximum Power Dissipation	160	W
$P_D @ T_C = 100^\circ C$	Maximum Power Dissipation	65	
$T_J$	Operating Junction and Storage Temperature Range	-55 to +150	$^\circ C$
$T_{STG}$			
	Mounting Torque, 6-32 or M3 Screw.	10 lbf•in (1.1 N•m)	

### Thermal Resistance

	Parameter	Min.	Typ.	Max.	Units
$R_{\theta JC}$	Junction-to-Case - IGBT	—	—	0.77	$^\circ C/W$
$R_{\theta JC}$	Junction-to-Case - Diode	—	—	1.7	
$R_{\theta CS}$	Case-to-Sink, flat, greased surface	—	0.24	—	
$R_{\theta JA}$	Junction-to-Ambient, typical socket mount	—	—	40	
Wt	Weight	—	6 (0.21)	—	g (oz)

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2/7/2000

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# IRG4PH40KD

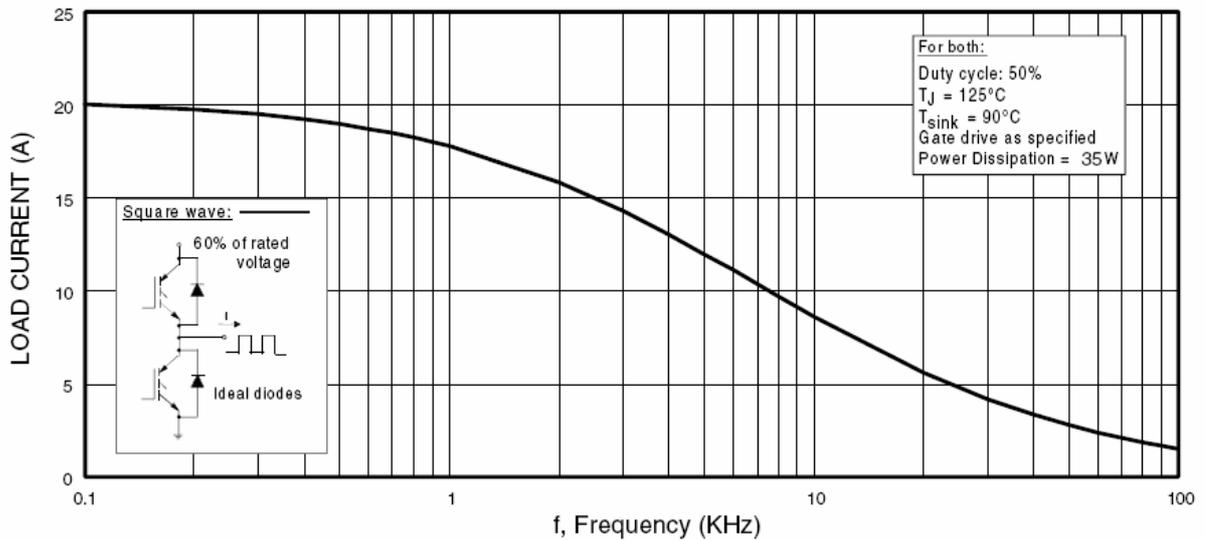
International  
**IRF** Rectifier

## Electrical Characteristics @ $T_J = 25^\circ\text{C}$ (unless otherwise specified)

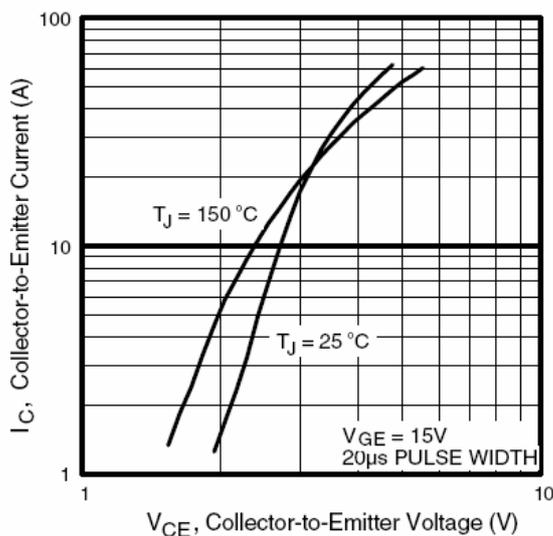
	Parameter	Min.	Typ.	Max.	Units	Conditions
$V_{(BR)CES}$	Collector-to-Emitter Breakdown Voltage <sup>③</sup>	1200	—	—	V	$V_{GE} = 0V, I_C = 250\mu A$
$\Delta V_{(BR)CES}/\Delta T_J$	Temperature Coeff. of Breakdown Voltage	—	0.37	—	V/ $^\circ\text{C}$	$V_{GE} = 0V, I_C = 1.0mA$
$V_{CE(on)}$	Collector-to-Emitter Saturation Voltage	—	2.74	3.4	V	$I_C = 15A$ $I_C = 30A$ $I_C = 15A, T_J = 150^\circ\text{C}$
		—	3.29	—		
		—	2.53	—		
$V_{GE(th)}$	Gate Threshold Voltage	3.0	—	6.0		$V_{CE} = V_{GE}, I_C = 250\mu A$
$\Delta V_{GE(th)}/\Delta T_J$	Temperature Coeff. of Threshold Voltage	—	-3.3	—	mV/ $^\circ\text{C}$	$V_{CE} = V_{GE}, I_C = 250\mu A$
$g_{fe}$	Forward Transconductance <sup>④</sup>	8.0	12	—	S	$V_{CE} = 100V, I_C = 15A$
$I_{CES}$	Zero Gate Voltage Collector Current	—	—	250	$\mu A$	$V_{GE} = 0V, V_{CE} = 1200V$ $V_{GE} = 0V, V_{CE} = 1200V, T_J = 150^\circ\text{C}$
		—	—	3000		
$V_{FM}$	Diode Forward Voltage Drop	—	2.6	3.3	V	$I_C = 8.0A$ $I_C = 8.0A, T_J = 125^\circ\text{C}$
		—	2.4	3.1		
$I_{GES}$	Gate-to-Emitter Leakage Current	—	—	$\pm 100$	nA	$V_{GE} = \pm 20V$

## Switching Characteristics @ $T_J = 25^\circ\text{C}$ (unless otherwise specified)

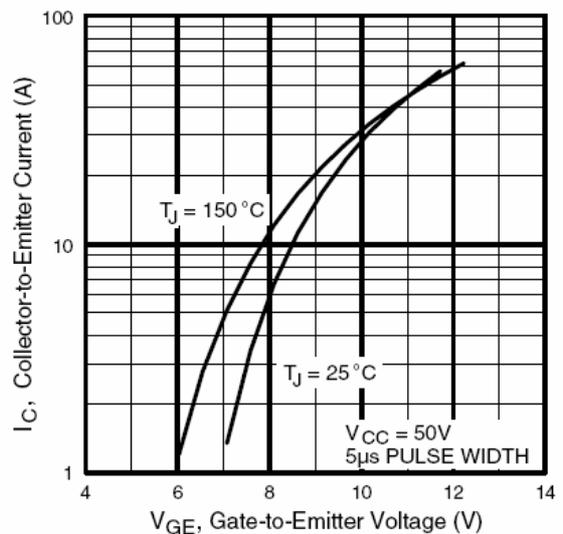
	Parameter	Min.	Typ.	Max.	Units	Conditions
$Q_g$	Total Gate Charge (turn-on)	—	94	140	nC	$I_C = 15A$ $V_{CC} = 400V$ $V_{GE} = 15V$
$Q_{ge}$	Gate - Emitter Charge (turn-on)	—	14	22		
$Q_{gc}$	Gate - Collector Charge (turn-on)	—	37	55		
$t_{d(on)}$	Turn-On Delay Time	—	50	—	ns	$T_J = 25^\circ\text{C}$ $I_C = 15A, V_{CC} = 800V$ $V_{GE} = 15V, R_G = 10\Omega$
$t_r$	Rise Time	—	31	—		
$t_{d(off)}$	Turn-Off Delay Time	—	96	140		
$t_f$	Fall Time	—	220	330		
$E_{on}$	Turn-On Switching Loss	—	1.31	—	mJ	Energy losses include "tail" and diode reverse recovery See Fig. 9,10,18
$E_{off}$	Turn-Off Switching Loss	—	1.12	—		
$E_{ts}$	Total Switching Loss	—	2.43	2.8		
$t_{sc}$	Short Circuit Withstand Time	10	—	—	$\mu s$	$V_{CC} = 720V, T_J = 125^\circ\text{C}$ $V_{GE} = 15V, R_G = 10\Omega, V_{CPK} < 500V$
$t_{d(on)}$	Turn-On Delay Time	—	49	—	ns	$T_J = 150^\circ\text{C}$ , $I_C = 15A, V_{CC} = 800V$ $V_{GE} = 15V, R_G = 10\Omega$ Energy losses include "tail" and diode reverse recovery
$t_r$	Rise Time	—	33	—		
$t_{d(off)}$	Turn-Off Delay Time	—	290	—		
$t_f$	Fall Time	—	440	—		
$E_{ts}$	Total Switching Loss	—	5.1	—	mJ	Measured 5mm from package
$L_E$	Internal Emitter Inductance	—	13	—	nH	Measured 5mm from package
$C_{ies}$	Input Capacitance	—	1600	—	pF	$V_{GE} = 0V$ $V_{CC} = 30V$ $f = 1.0MHz$
$C_{oes}$	Output Capacitance	—	77	—		
$C_{res}$	Reverse Transfer Capacitance	—	26	—		
$t_{rr}$	Diode Reverse Recovery Time	—	63	95	ns	$T_J = 25^\circ\text{C}$ See Fig. 14 $T_J = 125^\circ\text{C}$ 14
		—	106	160		
$I_{rr}$	Diode Peak Reverse Recovery Current	—	4.5	8.0	A	$T_J = 25^\circ\text{C}$ See Fig. 15 $T_J = 125^\circ\text{C}$ 15
		—	6.2	11		
$Q_{rr}$	Diode Reverse Recovery Charge	—	140	380	nC	$T_J = 25^\circ\text{C}$ See Fig. 16 $T_J = 125^\circ\text{C}$ 16
		—	335	880		
$di_{(rec)M}/dt$	Diode Peak Rate of Fall of Recovery During $t_b$	—	133	—	A/ $\mu s$	$T_J = 25^\circ\text{C}$ See Fig. 17 $T_J = 125^\circ\text{C}$ 17
		—	85	—		



**Fig. 1 - Typical Load Current vs. Frequency**  
(Load Current =  $I_{RMS}$  of fundamental)



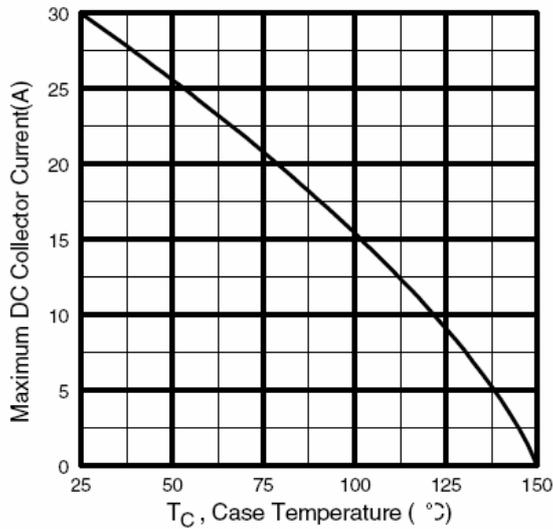
**Fig. 2 - Typical Output Characteristics**



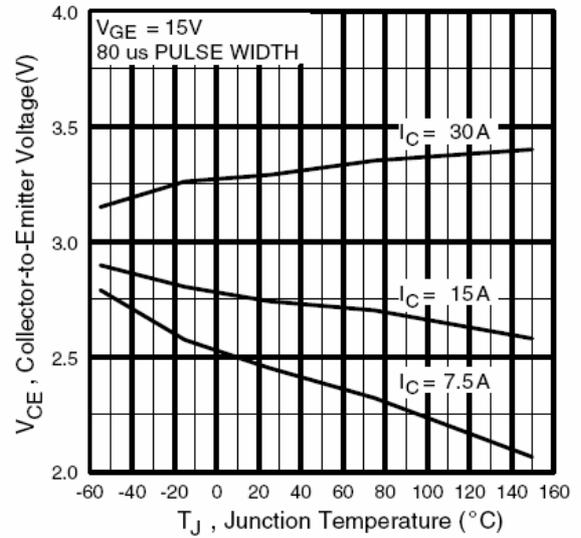
**Fig. 3 - Typical Transfer Characteristics**

# IRG4PH40KD

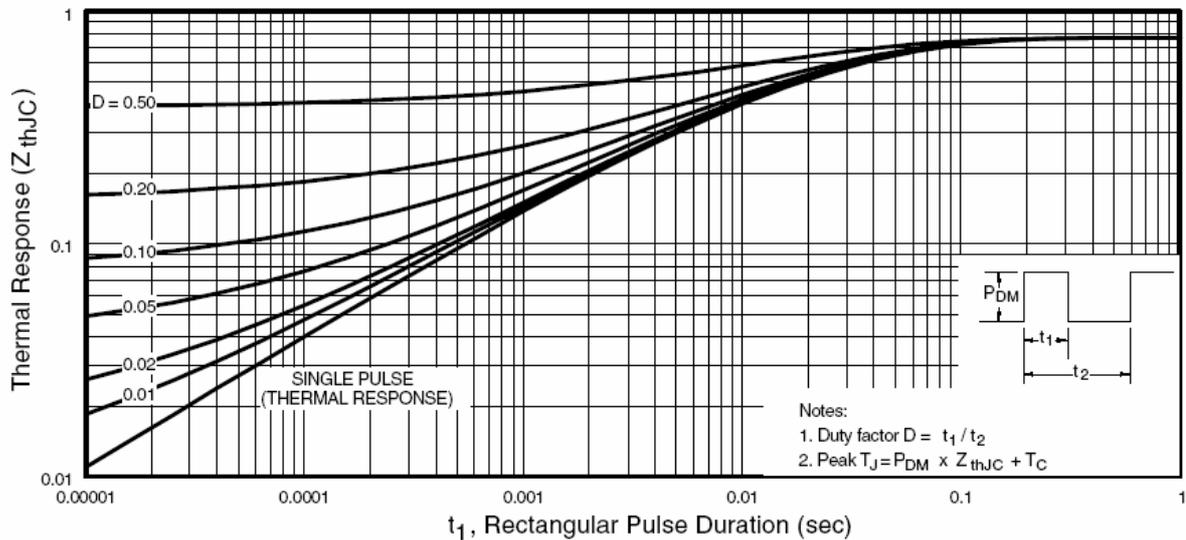
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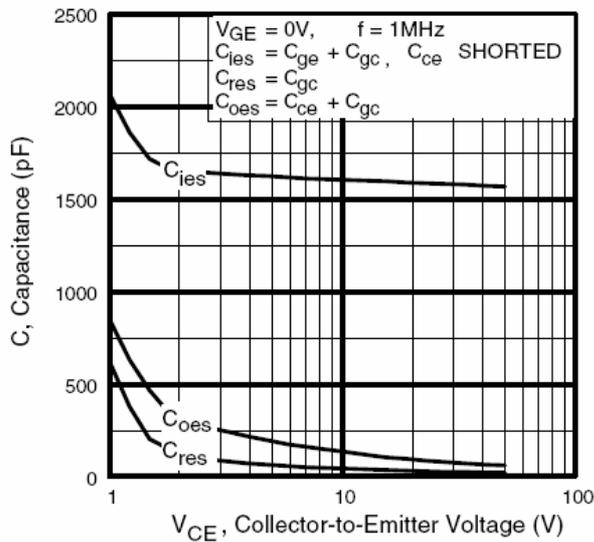
**Fig. 4** - Maximum Collector Current vs. Case Temperature



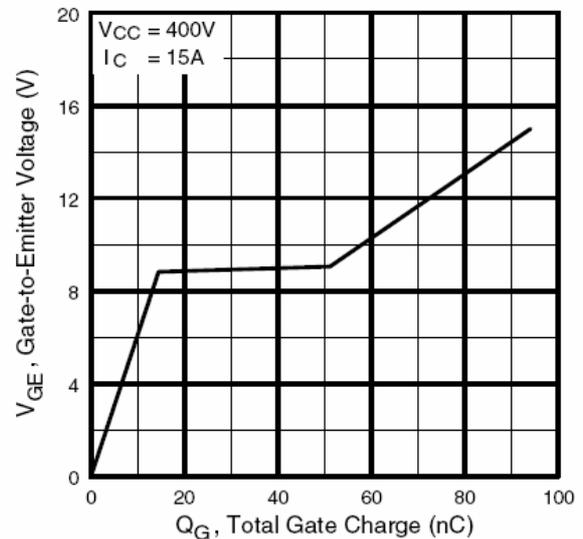
**Fig. 5** - Typical Collector-to-Emitter Voltage vs. Junction Temperature



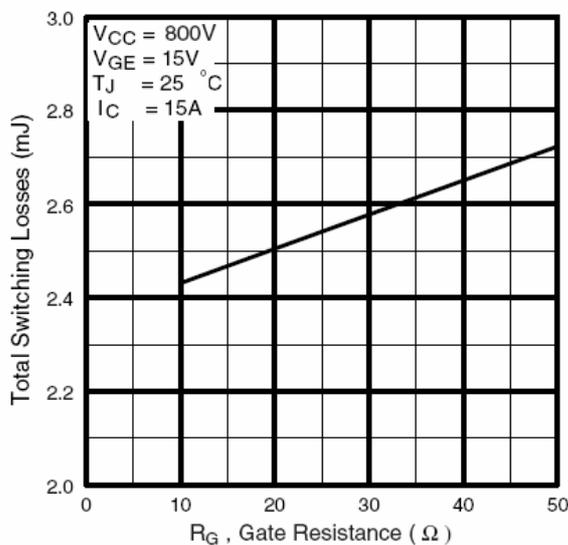
**Fig. 6** - Maximum Effective Transient Thermal Impedance, Junction-to-Case



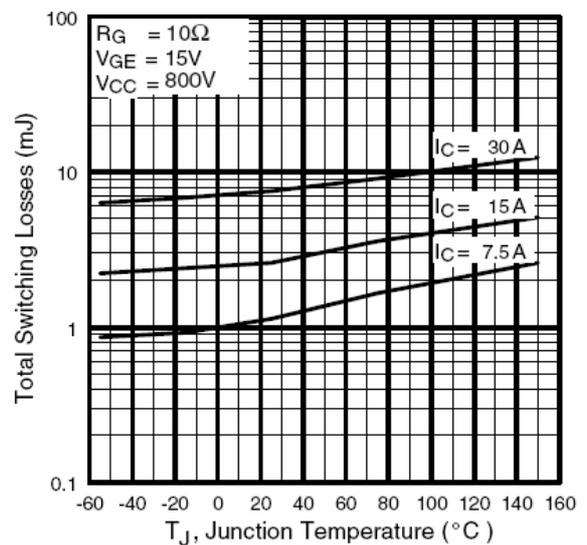
**Fig. 7** - Typical Capacitance vs. Collector-to-Emitter Voltage



**Fig. 8** - Typical Gate Charge vs. Gate-to-Emitter Voltage



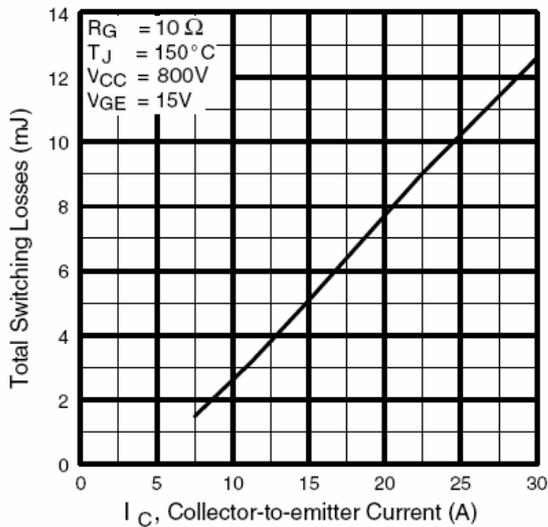
**Fig. 9** - Typical Switching Losses vs. Gate Resistance



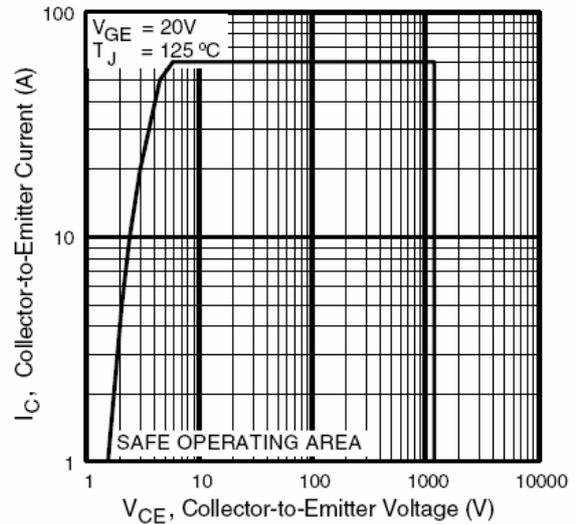
**Fig. 10** - Typical Switching Losses vs. Junction Temperature

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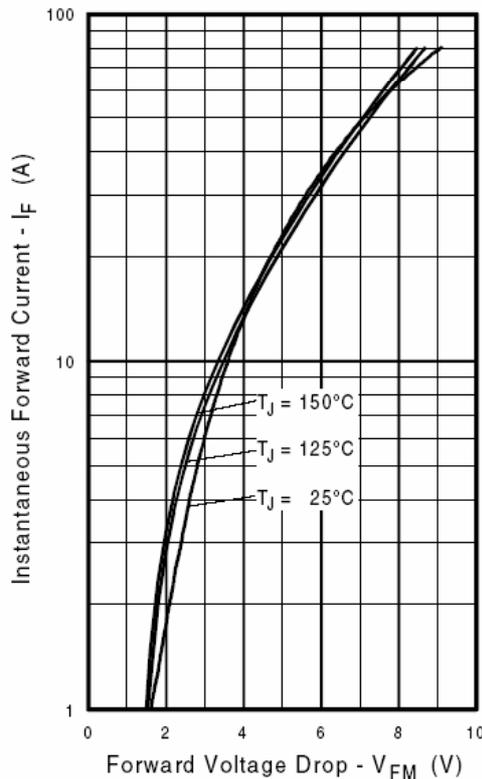
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**Fig. 11** - Typical Switching Losses vs. Collector-to-emitter Current



**Fig. 12** - Turn-Off SOA



**Fig. 13** - Maximum Forward Voltage Drop vs. Instantaneous Forward Current

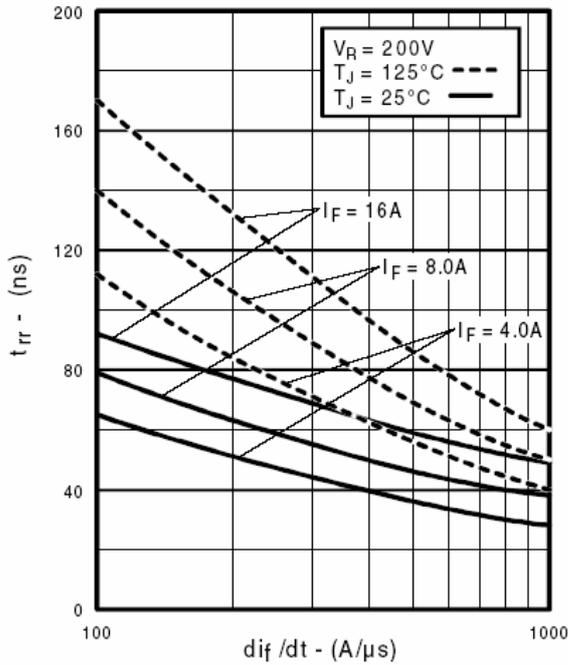


Fig. 14 - Typical Reverse Recovery vs.  $di_f/dt$

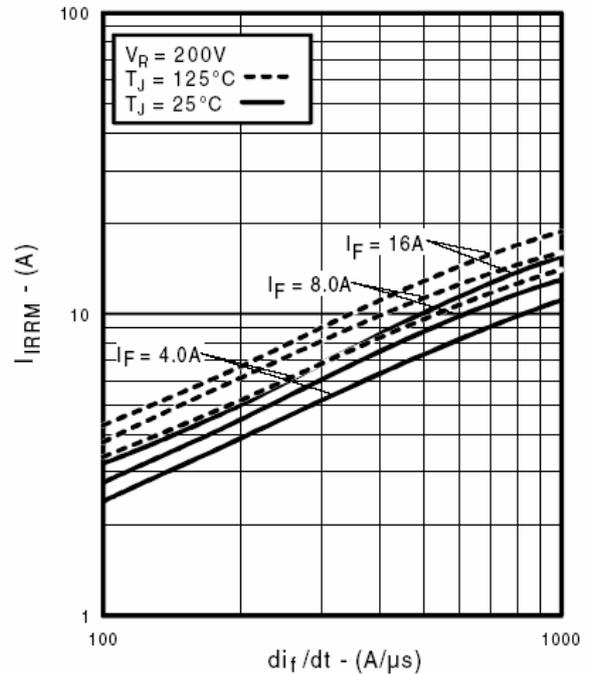


Fig. 15 - Typical Recovery Current vs.  $di_f/dt$

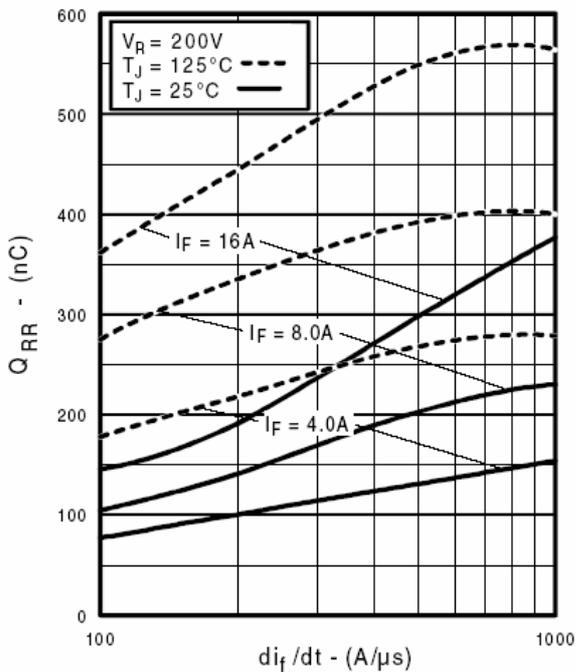


Fig. 16 - Typical Stored Charge vs.  $di_f/dt$

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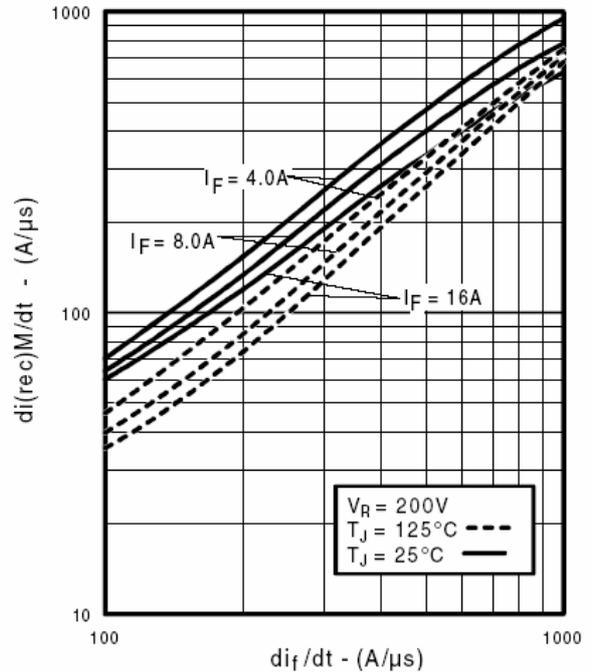
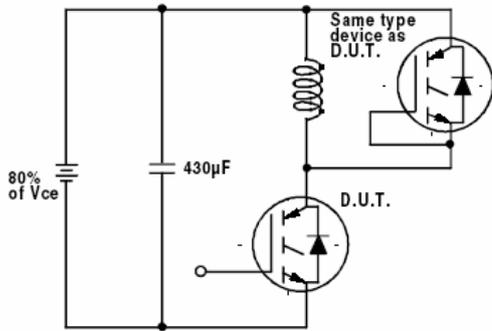
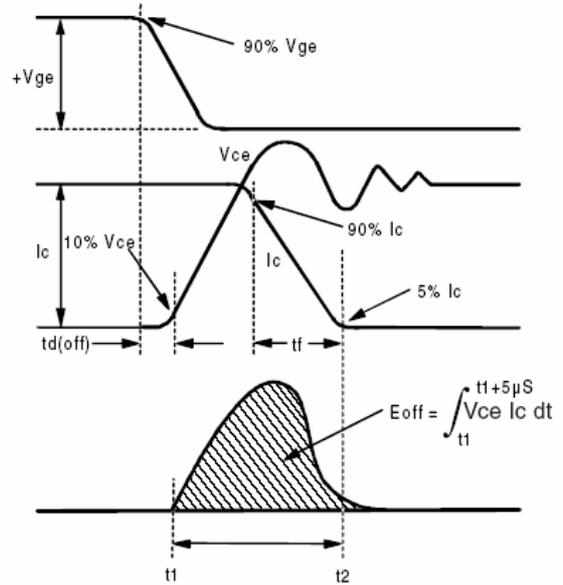


Fig. 17 - Typical  $di_{(rec)M}/dt$  vs.  $di_f/dt$

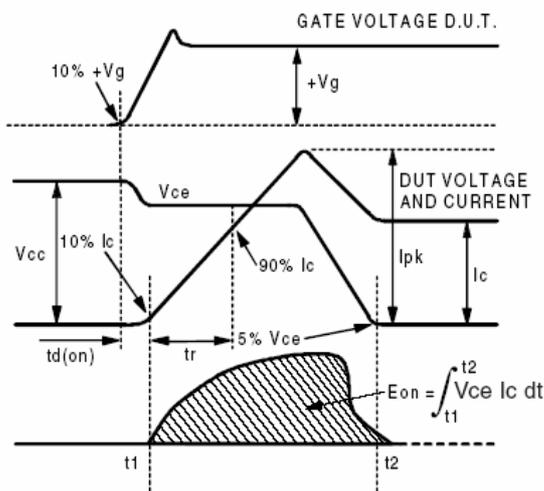
# IRG4PH40KD



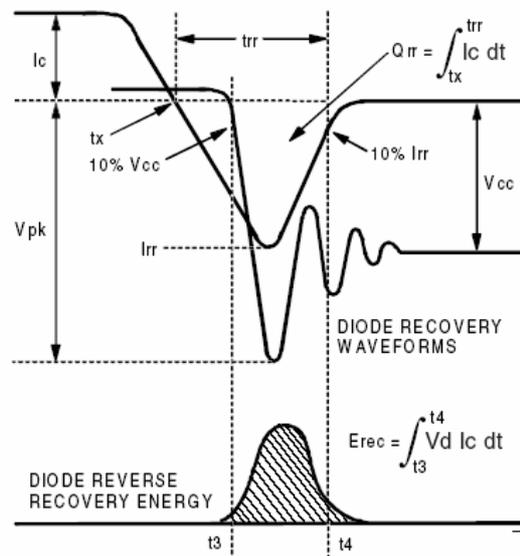
**Fig. 18a** - Test Circuit for Measurement of  $I_{LM}$ ,  $E_{on}$ ,  $E_{off}(\text{diode})$ ,  $t_{rr}$ ,  $Q_{rr}$ ,  $I_{rr}$ ,  $t_{d(on)}$ ,  $t_r$ ,  $t_{d(off)}$ ,  $t_f$



**Fig. 18b** - Test Waveforms for Circuit of Fig. 18a, Defining  $E_{off}$ ,  $t_{d(off)}$ ,  $t_f$



**Fig. 18c** - Test Waveforms for Circuit of Fig. 18a, Defining  $E_{on}$ ,  $t_{d(on)}$ ,  $t_r$



**Fig. 18d** - Test Waveforms for Circuit of Fig. 18a, Defining  $E_{rec}$ ,  $t_{rr}$ ,  $Q_{rr}$ ,  $I_{rr}$

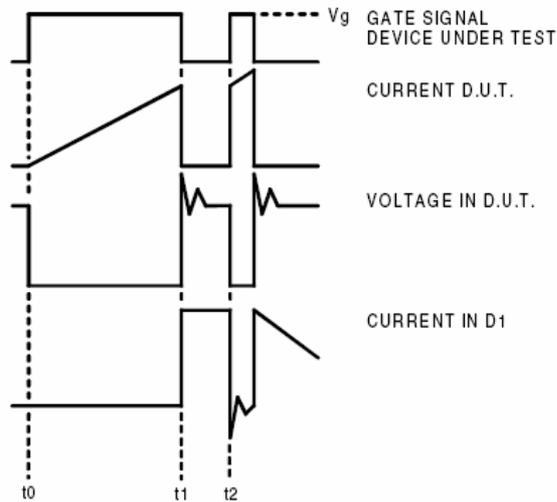


Figure 18e. Macro Waveforms for Figure 18a's Test Circuit

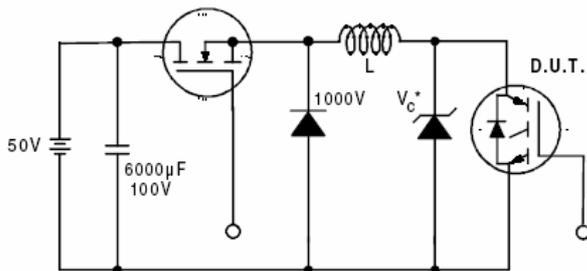


Figure 19. Clamped Inductive Load Test Circuit

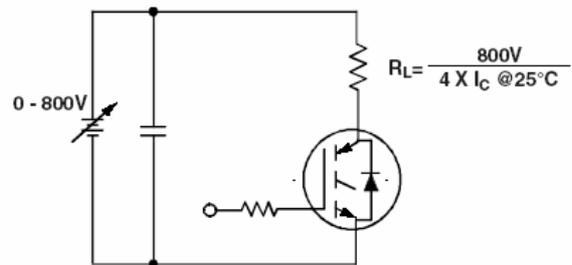


Figure 20. Pulsed Collector Current Test Circuit

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	Diseño e implementación de un convertidor monofásico de cinco niveles con control basado en DSP	José Francisco Campos Bizcocho

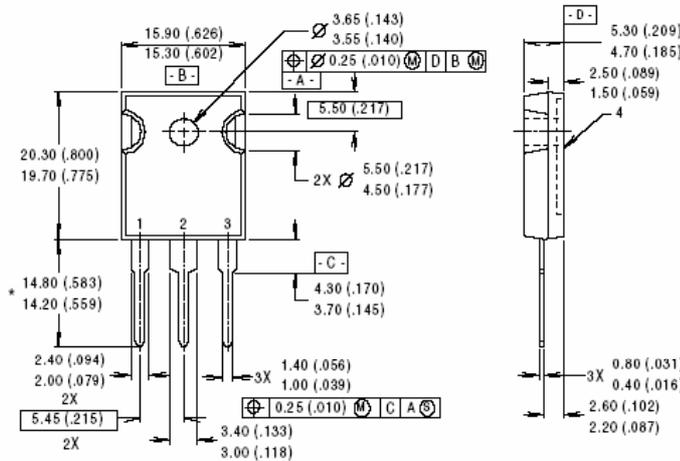
# IRG4PH40KD

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**IR** Rectifier

## Notes:

- ① Repetitive rating:  $V_{GE}=20V$ ; pulse width limited by maximum junction temperature (figure 20)
- ②  $V_{CC}=80\%(V_{CES})$ ,  $V_{GE}=20V$ ,  $L=10\mu H$ ,  $R_G=10\Omega$  (figure 19)
- ③ Pulse width  $\leq 80\mu s$ ; duty factor  $\leq 0.1\%$ .
- ④ Pulse width  $5.0\mu s$ , single shot.

## Case Outline — TO-247AC



### NOTES:

- 1 DIMENSIONS & TOLERANCING PER ANSI Y14.5M, 1982.
- 2 CONTROLLING DIMENSION: INCH.
- 3 DIMENSIONS ARE SHOWN MILLIMETERS (INCHES).
- 4 CONFORMS TO JEDEC OUTLINE TO-247AC.

### LEAD ASSIGNMENTS

- 1 - GATE
- 2 - COLLECTOR
- 3 - EMITTER
- 4 - COLLECTOR

\* LONGER LEADED (20mm) VERSION AVAILABLE (TO-247AD) TO ORDER ADD "E" SUFFIX TO PART NUMBER

**CONFORMS TO JEDEC OUTLINE TO-247AC (TO-3P)**  
Dimensions in Millimeters and (Inches)

International  
**IR** Rectifier

**IR WORLD HEADQUARTERS:** 233 Kansas St., El Segundo, California 90245, USA Tel: (310) 252-7105  
**IR EUROPEAN REGIONAL CENTRE:** 439/445 Godstone Rd, Whyteleafe, Surrey CR3 OBL, UK Tel: ++ 44 (0)20 8645 8000  
**IR CANADA:** 15 Lincoln Court, Brampton, Ontario L6T3Z2, Tel: (905) 453 2200  
**IR GERMANY:** Saalburgstrasse 157, 61350 Bad Homburg Tel: ++ 49 (0) 6172 96590  
**IR ITALY:** Via Liguria 49, 10071 Borgaro, Torino Tel: ++ 39 011 451 0111  
**IR JAPAN:** K&H Bldg., 2F, 30-4 Nishi-Ikebukuro 3-Chome, Toshima-Ku, Tokyo 171 Tel: 81 (0)3 3983 0086  
**IR SOUTHEAST ASIA:** 1 Kim Seng Promenade, Great World City West Tower, 13-11, Singapore 237994 Tel: ++ 65 (0)838 4630  
**IR TAIWAN:** 16 Fl. Suite D. 207, Sec. 2, Tun Haw South Road, Taipei, 10673 Tel: 886-(0)2 2377 9936  
*Data and specifications subject to change without notice. 6/00*

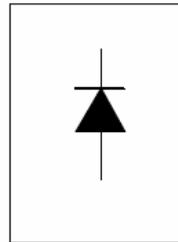
## 2. 10ETF10.

I2146 rev. A 11/99

International  
**IOR** Rectifier

**QUIETIR** Series  
10ETF..

**FAST SOFT RECOVERY  
RECTIFIER DIODE**



$V_F < 1.33V @ 10A$   
 $t_{rr} = 80ns$   
 $V_{RRM} 1000 \text{ to } 1200V$

### Description/Features

The 10ETF.. fast soft recovery **QUIETIR** rectifier series has been optimized for combined short reverse recovery time and low forward voltage drop.

The glass passivation ensures stable reliable operation in the most severe temperature and power cycling conditions.

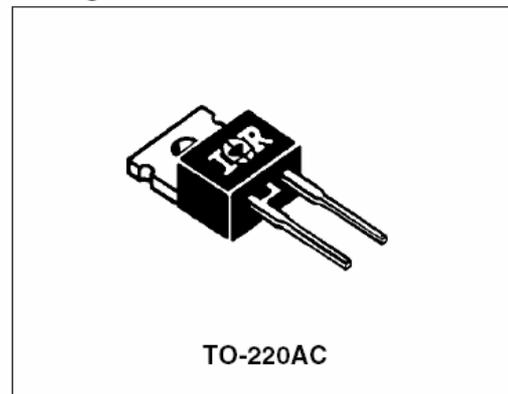
Typical applications are both:

- output rectification and freewheeling in inverters, choppers and converters
- and input rectifications where severe restrictions on conducted EMI should be met.

### Major Ratings and Characteristics

Characteristics	10ETF..	Units
$I_{F(AV)}$ Sinusoidal waveform	10	A
$V_{RRM}$	1000to 1200	V
$I_{FSM}$	160	A
$V_F @ 10A, T_J=25^\circ C$	1.33	V
$t_{rr} @ 1A, 100A/\mu s$	80	ns
$T_J$	-40to 150	$^\circ C$

### Package Outline



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## 10ETF.. QUIETIR Series

I2146 rev. A 11/99

International  
**IR** Rectifier

### Voltage Ratings

Part Number	$V_{RRM}$ , maximum peak reverse voltage V	$V_{RSM}$ , maximum non repetitive peak reverse voltage V	$I_{FRM}$ 150°C mA
10ETF10	1000	1100	4
10ETF12	1200	1300	

### Absolute Maximum Ratings

Parameters	10ETF..	Units	Conditions
$I_{F(AV)}$ Max. Average Forward Current	10	A	@ $T_C = 125^\circ\text{C}$ , 180° conduction half sine wave
$I_{FSM}$ Max. Peak One Cycle Non-Repetitive Surge Current	160	A	10ms Sine pulse, rated $V_{RRM}$ applied
	185		10ms Sine pulse, no voltage reapplied
$I^2t$ Max. $I^2t$ for fusing	128	$\text{A}^2\text{s}$	10ms Sine pulse, rated $V_{RRM}$ applied
	180		10ms Sine pulse, no voltage reapplied
$I^2\sqrt{t}$ Max. $I^2\sqrt{t}$ for fusing	1800	$\text{A}^2\sqrt{\text{s}}$	$t = 0.1$ to 10ms, no voltage reapplied

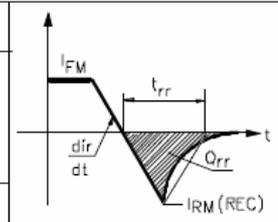
### Electrical Specifications

Parameters	10ETF..	Units	Conditions
$V_{FM}$ Max. Forward Voltage Drop	1.33	V	@ 10A, $T_J = 25^\circ\text{C}$
$r_t$ Forward slope resistance	22.9	$\text{m}\Omega$	$T_J = 150^\circ\text{C}$
$V_{F(TO)}$ Threshold voltage	0.96	V	
$I_{FRM}$ Max. Reverse Leakage Current	0.1	mA	$T_J = 25^\circ\text{C}$
	4		$T_J = 150^\circ\text{C}$

$V_R = \text{rated } V_{RRM}$

### Recovery Characteristics

Parameters	10ETF..	Units	Conditions
$t_{rr}$ Reverse Recovery Time	310	ns	$I_F @ 10\text{Apk}$ @ 25A/ $\mu\text{s}$ @ 25°C
$I_{rr}$ Reverse Recovery Current	4.7	A	
$Q_{rr}$ Reverse Recovery Charge	1.05	$\mu\text{C}$	
S Typical Snap Factor	0.6		



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### Thermal-Mechanical Specifications

Parameters	10ETF..	Units	Conditions
$T_J$ Max. Junction Temperature Range	-40 to 150	°C	
$T_{stg}$ Max. Storage Temperature Range	-40 to 150	°C	
$R_{thJC}$ Max. Thermal Resistance Junction to Case	1.5	°C/W	DC operation
$R_{thJA}$ Max. Thermal Resistance Junction to Ambient	62	°C/W	
$R_{thCS}$ Typical Thermal Resistance, Case to Heatsink	0.5	°C/W	Mounting surface, smooth and greased
wt Approximate Weight	2 (0.07)	g (oz.)	
T Mounting Torque	Min.	6 (5)	Kg-cm (lbf-in)
	Max.	12 (10)	
Case Style	TO-220AC		JEDEC

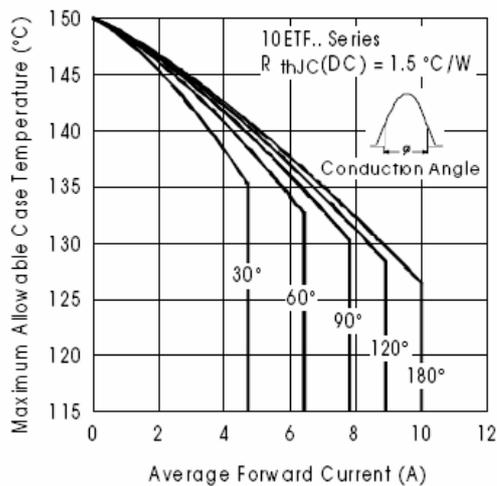


Fig. 1 - Current Rating Characteristics

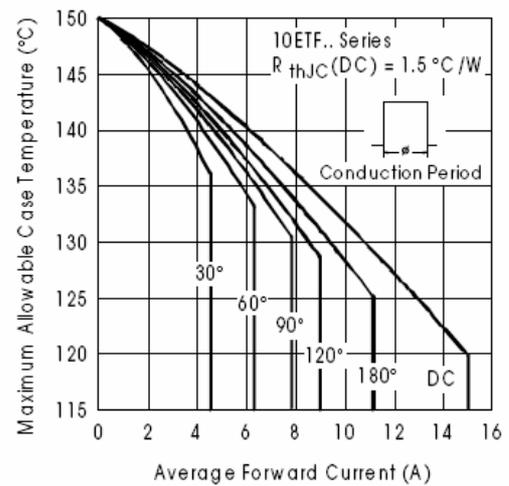


Fig. 2 - Current Rating Characteristics

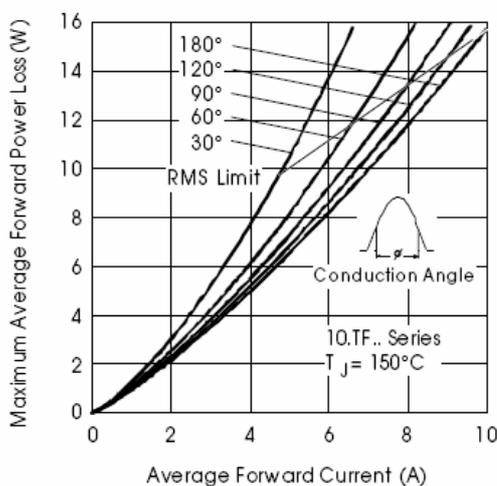


Fig. 3 - Forward Power Loss Characteristics

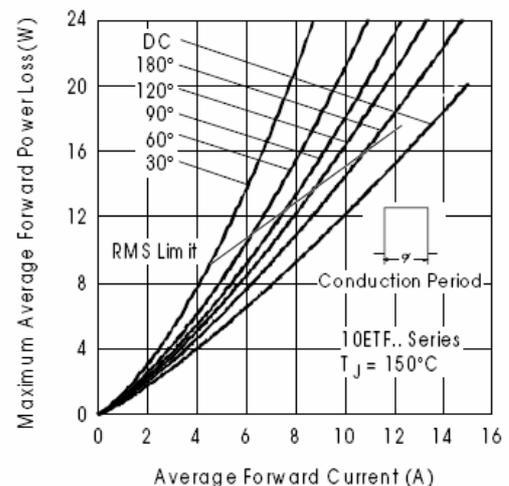


Fig. 4 - Forward Power Loss Characteristics

# 10ETF.. QUIETIR Series

I2146 rev. A 11/99

International  
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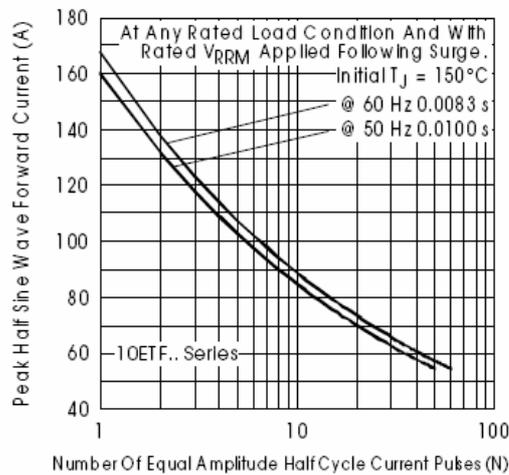


Fig. 5 - Maximum Non-Repetitive Surge Current

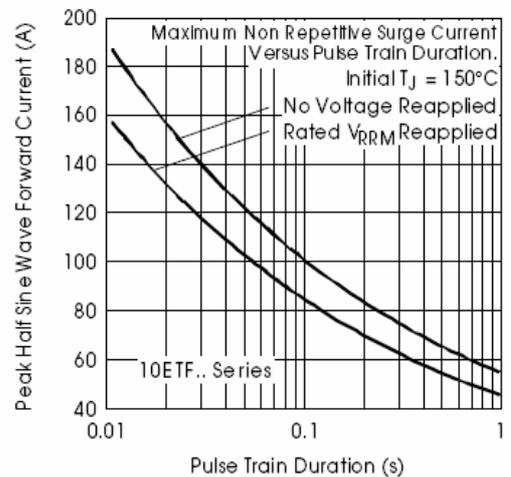


Fig. 6 - Maximum Non-Repetitive Surge Current

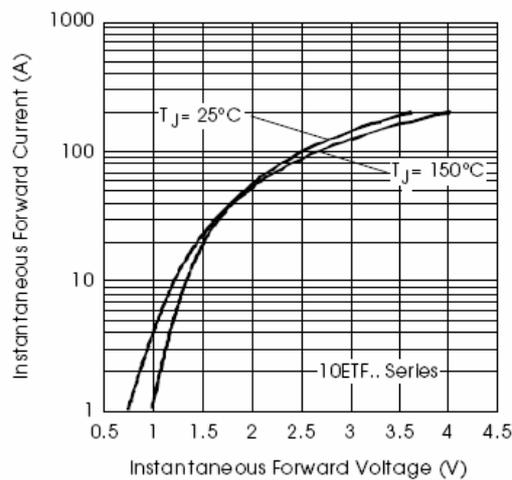


Fig. 7 - Forward Voltage Drop Characteristics

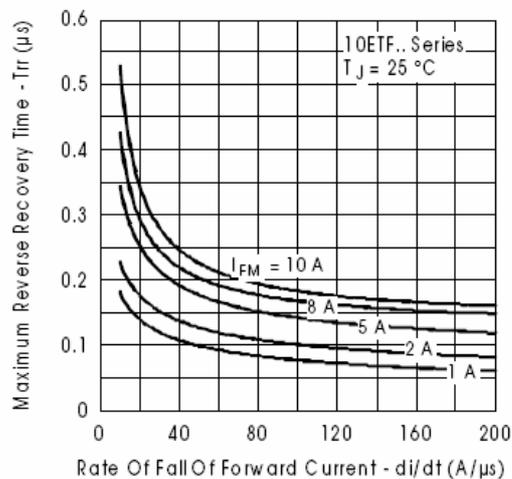


Fig. 8 - Recovery Time Characteristics,  $T_J = 25^\circ\text{C}$

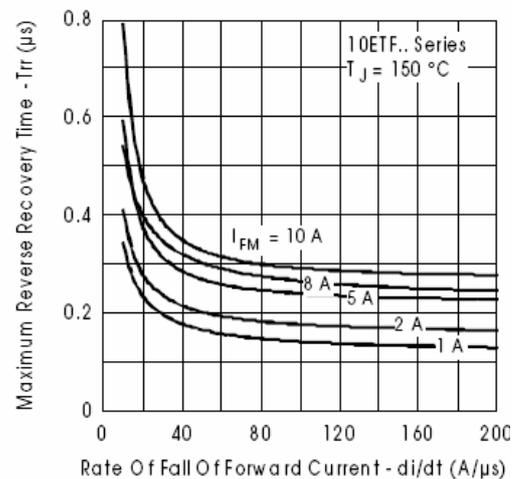


Fig. 9 - Recovery Time Characteristics,  $T_J = 150^\circ\text{C}$

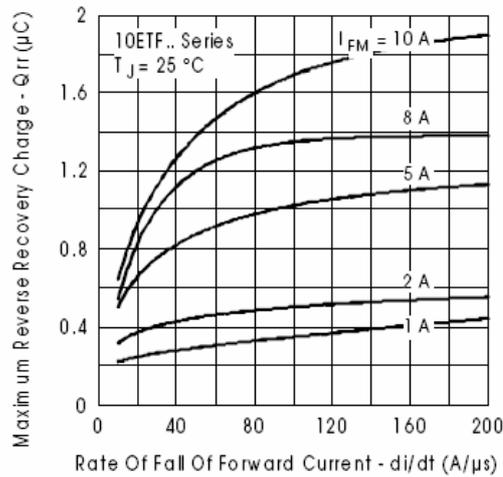


Fig. 10 - Recovery Charge Characteristics,  $T_J = 25^\circ\text{C}$

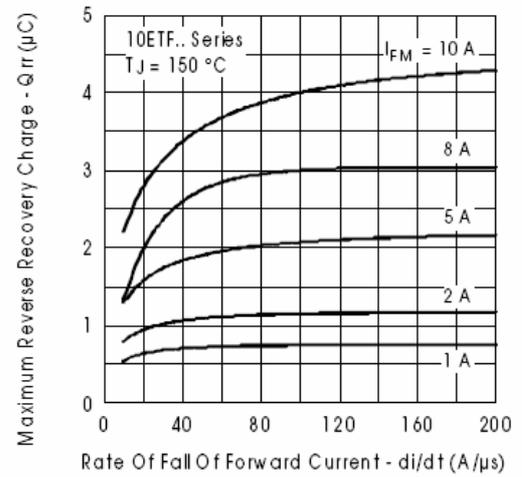


Fig. 11 - Recovery Charge Characteristics,  $T_J = 150^\circ\text{C}$

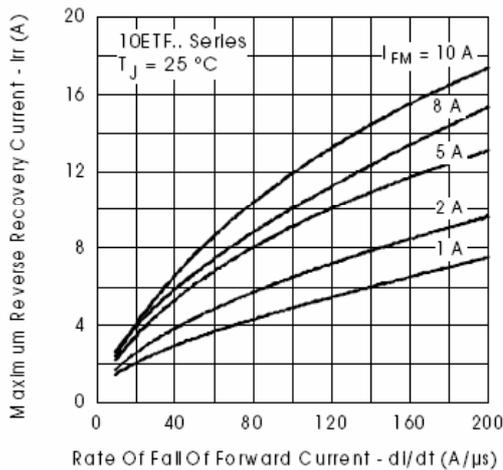


Fig. 12 - Recovery Current Characteristics,  $T_J = 25^\circ\text{C}$

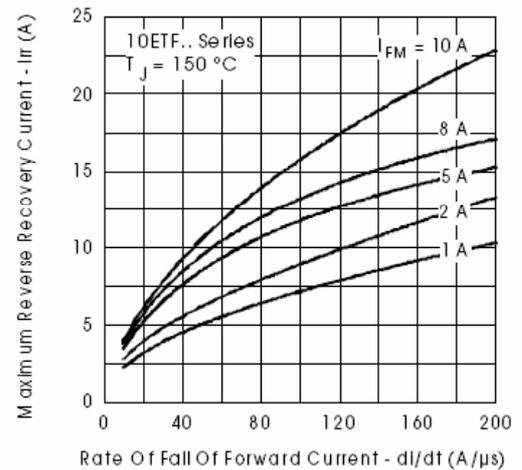


Fig. 13 - Recovery Current Characteristics,  $T_J = 150^\circ\text{C}$

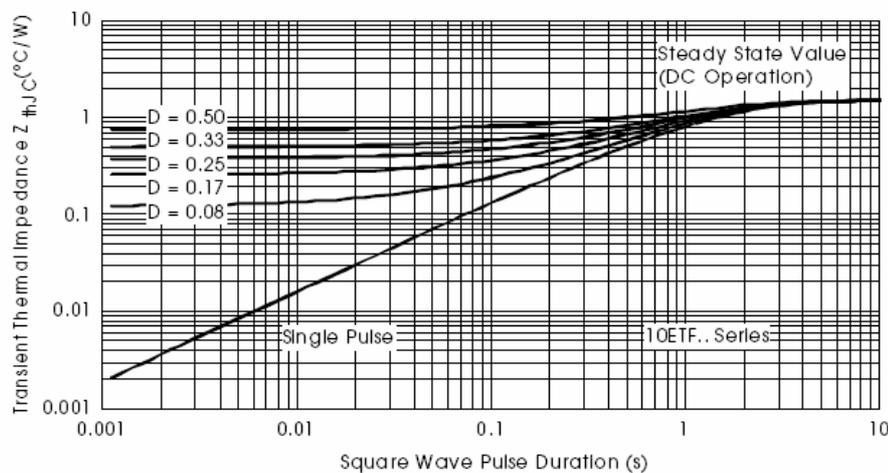


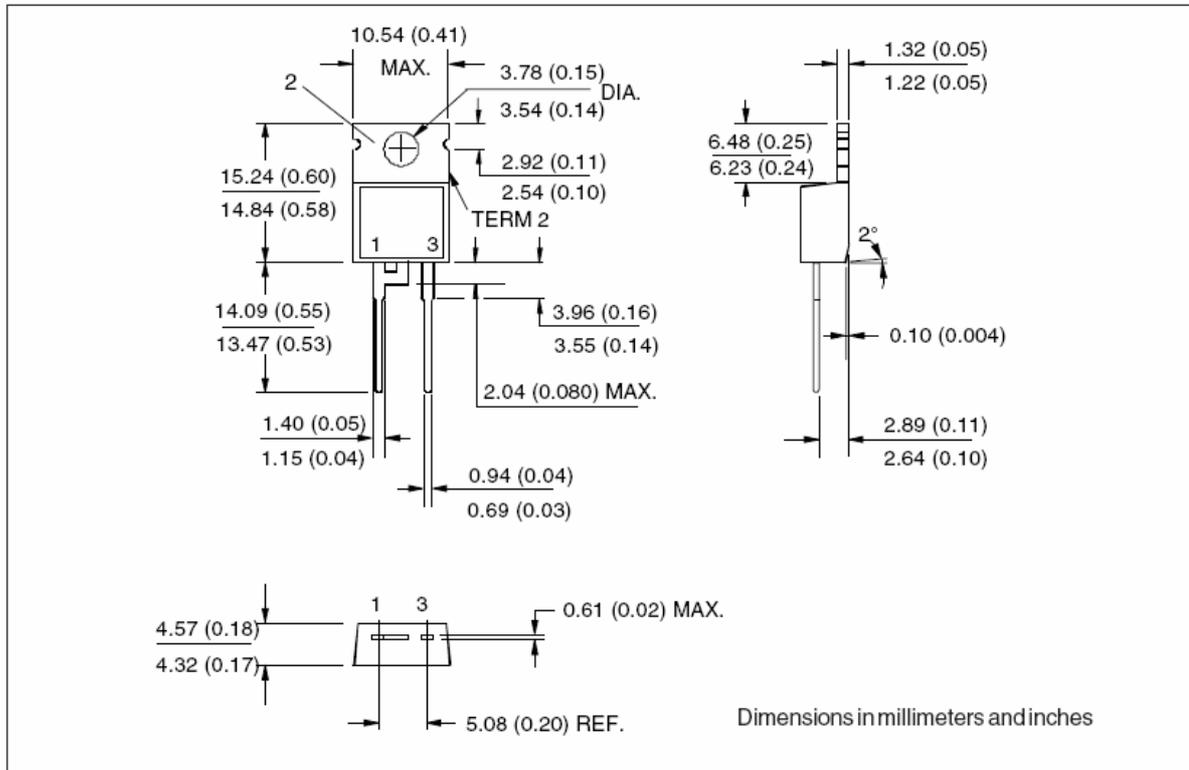
Fig. 14 - Thermal Impedance  $Z_{thJC}$  Characteristics

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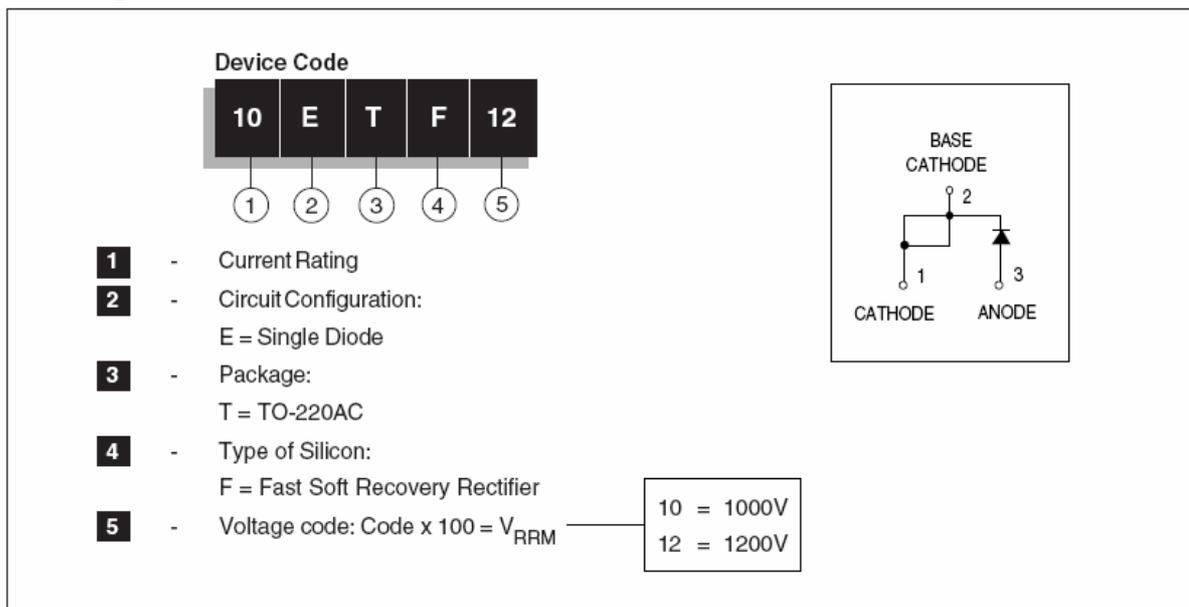
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## Outline Table



## Ordering Information Table



	Proyecto Fin de Carrera	Alumno
	Diseño e implementación de un convertidor monofásico de cinco niveles con control basado en DSP	José Francisco Campos Bizcocho

International  
**IOR** Rectifier

10ETF.. QUIET**IR** Series  
I2146 rev. A 11/99

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**WORLD HEADQUARTERS:** 233 Kansas St., El Segundo, California 90245 U.S.A. Tel: (310) 322 3331. Fax: (310) 322 3332.  
**EUROPEAN HEADQUARTERS:** Hurst Green, Oxted, Surrey RH8 9BB, U.K. Tel: ++ 44 1883 732020. Fax: ++ 44 1883 733408.  
**IR CANADA:** 15 Lincoln Court, Brampton, Markham, Ontario L6T3Z2. Tel: (905) 453 2200. Fax: (905) 475 8801.  
**IR GERMANY:** Saalburgstrasse 157, 61350 Bad Homburg. Tel: ++ 49 6172 96590. Fax: ++ 49 6172 965933.  
**IR ITALY:** Via Liguria 49, 10071 Borgaro, Torino. Tel: ++ 39 11 4510111. Fax: ++ 39 11 4510220.  
**IR FAR EAST:** K&H Bldg., 2F, 30-4 Nishi-Ikebukuro 3-Chome, Toshima-Ku, Tokyo, Japan 171. Tel: 81 3 3983 0086.  
**IR SOUTHEAST ASIA:** 1 Kim Seng Promenade, Great World City West Tower, 13-11, Singapore 237994. Tel: ++ 65 838 4630.  
**IR TAIWAN:** 16 Fl. Suite D.207, Sec. 2, Tun Haw South Road, Taipei, 10673, Taiwan. Tel: 886 2 2377 9936.

<http://www.irf.com>

Fax-On-Demand: +44 1883 733420

Data and specifications subject to change without notice.

### 3. IR2118.

International  
**IR** Rectifier

Data Sheet No. PD60146 Rev N

## IR2117(S)/IR2118(S) & (PbF)

### SINGLE CHANNEL DRIVER

#### Features

- Floating channel designed for bootstrap operation Fully operational to +600V Tolerant to negative transient voltage dV/dt immune
- Gate drive supply range from 10 to 20V
- Undervoltage lockout
- CMOS Schmitt-triggered inputs with pull-down
- Output in phase with input (IR2117) or out of phase with input (IR2118)
- Also available LEAD-FREE

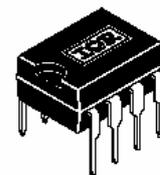
#### Description

The IR2117/IR2118(S) is a high voltage, high speed power MOSFET and IGBT driver. Proprietary HVIC and latch immune CMOS technologies enable ruggedized monolithic construction. The logic input is compatible with standard CMOS outputs. The output driver features a high pulse current buffer stage designed for minimum cross-conduction. The floating channel can be used to drive an N-channel power MOSFET or IGBT in the high or low side configuration which operates up to 600 volts.

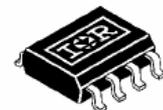
#### Product Summary

$V_{\text{OFFSET}}$	600V max.
$I_{\text{O}+/-}$	200 mA / 420 mA
$V_{\text{OUT}}$	10 - 20V
$t_{\text{on/off}}$ (typ.)	125 & 105 ns

#### Packages

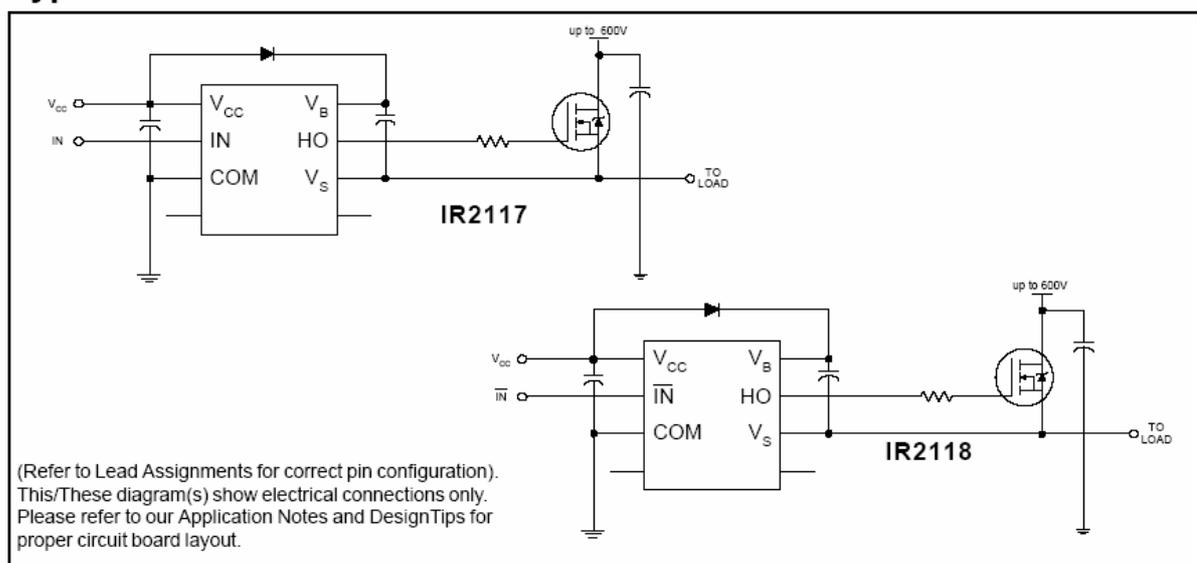


8-Lead PDIP  
IR2117/IR2118



8-Lead SOIC  
IR2117S/IR2118S

#### Typical Connection



	Proyecto Fin de Carrera	Alumno
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## IR2117(S)/IR2118(S) & (PbF)

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### Absolute Maximum Ratings

Absolute maximum ratings indicate sustained limits beyond which damage to the device may occur. All voltage parameters are absolute voltages referenced to COM. The thermal resistance and power dissipation ratings are measured under board mounted and still air conditions. Additional information is shown in Figures 5 through 8.

Symbol	Definition	Min.	Max.	Units	
$V_B$	High side floating supply voltage	-0.3	625	V	
$V_S$	High side floating supply offset voltage	$V_B - 25$	$V_B + 0.3$		
$V_{HO}$	High side floating output voltage	$V_S - 0.3$	$V_B + 0.3$		
$V_{CC}$	Logic supply voltage	-0.3	25		
$V_{IN}$	Logic input voltage	-0.3	$V_{CC} + 0.3$		
$dV_S/dt$	Allowable offset supply voltage transient (figure 2)	—	50	V/ns	
$P_D$	Package power dissipation @ $T_A \leq +25^\circ\text{C}$	(8 lead PDIP)	—	1.0	W
		(8 lead SOIC)	—	0.625	
$R_{thJA}$	Thermal resistance, junction to ambient	(8 lead PDIP)	—	125	$^\circ\text{C/W}$
		(8 lead SOIC)	—	200	
$T_J$	Junction temperature	—	150	$^\circ\text{C}$	
$T_S$	Storage temperature	-55	150		
$T_L$	Lead temperature (soldering, 10 seconds)	—	300		

### Recommended Operating Conditions

The input/output logic timing diagram is shown in figure 1. For proper operation the device should be used within the recommended conditions. The  $V_S$  offset rating is tested with all supplies biased at 15V differential.

Symbol	Definition	Min.	Max.	Units
$V_B$	High side floating supply absolute voltage	$V_S + 10$	$V_S + 20$	V
$V_S$	High side floating supply offset voltage	Note 1	600	
$V_{HO}$	High side floating output voltage	$V_S$	$V_B$	
$V_{CC}$	Logic supply voltage	10	20	
$V_{IN}$	Logic input voltage	0	$V_{CC}$	
$T_A$	Ambient temperature	-40	125	$^\circ\text{C}$

Note 1: Logic operational for  $V_S$  of -5 to +600V. Logic state held for  $V_S$  of -5V to  $-V_{BS}$ . (Please refer to the Design Tip DT97-3 for more details).

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## Dynamic Electrical Characteristics

$V_{BIAS}$  ( $V_{CC}$ ,  $V_{BS}$ ) = 15V,  $C_L$  = 1000 pF and  $T_A$  = 25°C unless otherwise specified. The dynamic electrical characteristics are measured using the test circuit shown in Figure 3.

Symbol	Definition	Min.	Typ.	Max.	Units	Test Conditions
$t_{on}$	Turn-on propagation delay	—	125	200	ns	$V_S = 0V$
$t_{off}$	Turn-off propagation delay	—	105	180		$V_S = 600V$
$t_r$	Turn-on rise time	—	80	130		
$t_f$	Turn-off fall time	—	40	65		

## Static Electrical Characteristics

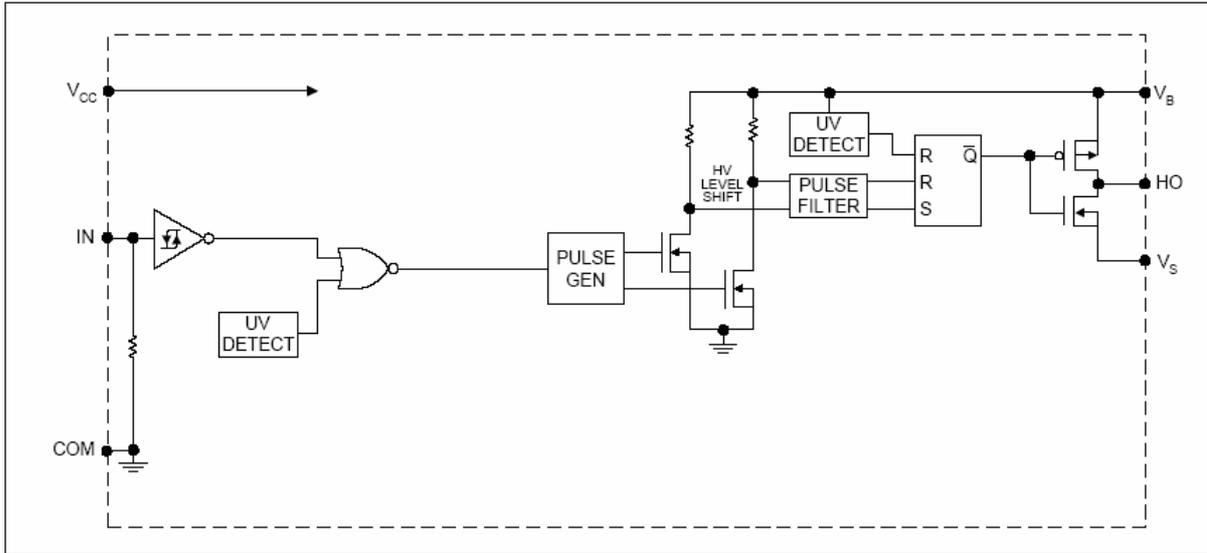
$V_{BIAS}$  ( $V_{CC}$ ,  $V_{BS}$ ) = 15V and  $T_A$  = 25°C unless otherwise specified. The  $V_{IN}$ ,  $V_{TH}$  and  $I_{IN}$  parameters are referenced to COM. The  $V_O$  and  $I_O$  parameters are referenced to COM and are applicable to the respective output leads: HO or LO.

Symbol	Definition	Min.	Typ.	Max.	Units	Test Conditions
$V_{IH}$	input voltage - logic "1" (IR2117) logic "0" (IR2118)	9.5	—	—	V	
$V_{IL}$	Input voltage - logic "0" (IR2117) logic "1" (IR2118)	—	—	6.0		
$V_{OH}$	High level output voltage, $V_{BIAS} - V_O$	—	—	100	mV	$I_O = 0A$
$V_{OL}$	Low level output voltage, $V_O$	—	—	100		$I_O = 0A$
$I_{LK}$	Offset supply leakage current	—	—	50	$\mu A$	$V_B = V_S = 600V$
$I_{QBS}$	Quiescent $V_{BS}$ supply current	—	50	240		$V_{IN} = 0V$ or $V_{CC}$
$I_{QCC}$	Quiescent $V_{CC}$ Supply Current	—	70	340		$V_{IN} = 0V$ or $V_{CC}$
$I_{IN+}$	Logic "1" input bias current (IR2117) (IR2118)	—	20	40		$V_{IN} = V_{CC}$
$I_{IN-}$	Logic "0" input bias current (IR2117) (IR2118)	—	—	1.0		$V_{IN} = 0V$ $V_{IN} = V_{CC}$
$V_{BSUV+}$	$V_{BS}$ supply undervoltage positive going threshold	7.6	8.6	9.6	V	
$V_{BSUV-}$	$V_{BS}$ supply undervoltage negative going threshold	7.2	8.2	9.2		
$V_{CCUV+}$	$V_{CC}$ supply undervoltage positive going threshold	7.6	8.6	9.6		
$V_{CCUV-}$	$V_{CC}$ supply undervoltage negative going threshold	7.2	8.2	9.2		
$I_{O+}$	Output high short circuit pulsed current	200	250	—	mA	$V_O = 0V$ $V_{IN} = \text{Logic "1"}$ $PW \leq 10 \mu s$
$I_{O-}$	Output low short circuit pulsed current	420	500	—		$V_O = 15V$ $V_{IN} = \text{Logic "0"}$ $PW \leq 10 \mu s$

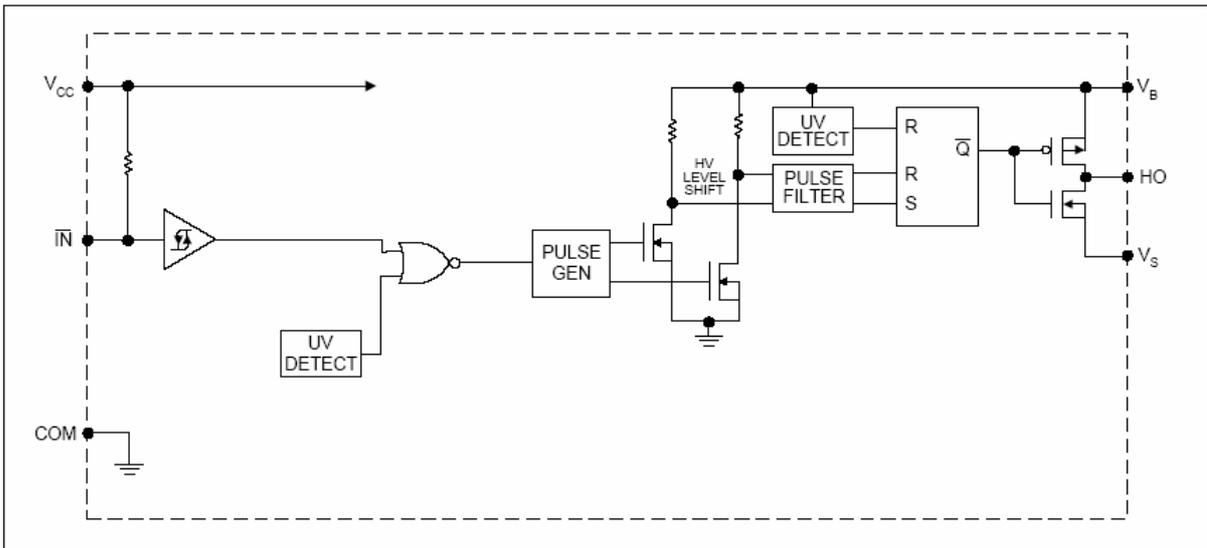
# IR2117(S)/IR2118(S) & (PbF)

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## Functional Block Diagram (IR2117)



## Functional Block Diagram (IR2118)



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## IR2117(S)/IR2118(S) & (PbF)

### Lead Definitions

Symbol	Description
V <sub>CC</sub>	Logic and gate drive supply
IN	Logic input for gate driver output (HO), in phase with HO (IR2117)
$\overline{\text{IN}}$	Logic input for gate driver output (HO), out of phase with HO (IR2118)
COM	Logic ground
V <sub>B</sub>	High side floating supply
HO	High side gate drive output
V <sub>S</sub>	High side floating supply return

### Lead Assignments

<p>8 Lead PDIP</p> <p><b>IR2117</b></p>	<p>8 Lead SOIC</p> <p><b>IR2117S</b></p>
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<p>8 Lead PDIP</p> <p><b>IR2118</b></p>	<p>8 Lead SOIC</p> <p><b>IR2118S</b></p>
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# IR2117(S)/IR2118(S) & (PbF)

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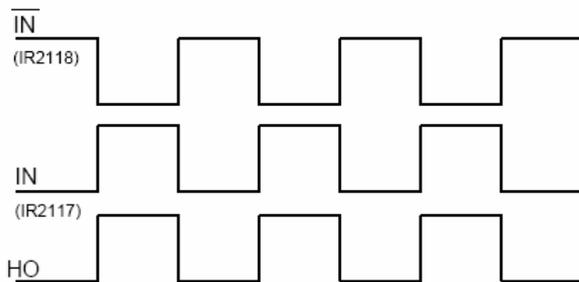


Figure 1. Input/Output Timing Diagram

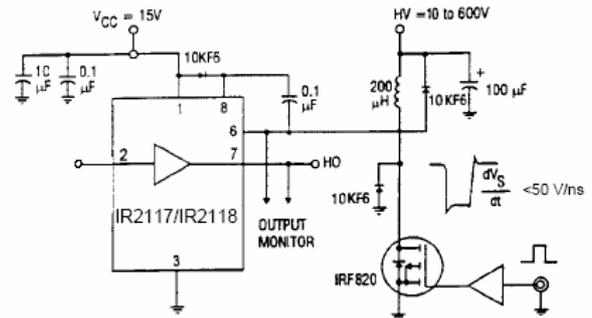


Figure 2. Floating Supply Voltage Transient Test Circuit

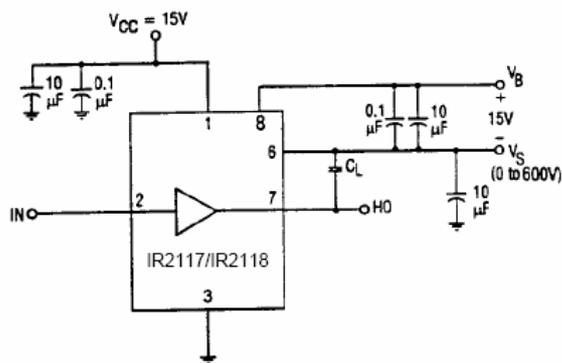


Figure 3. Switching Time Test Circuit

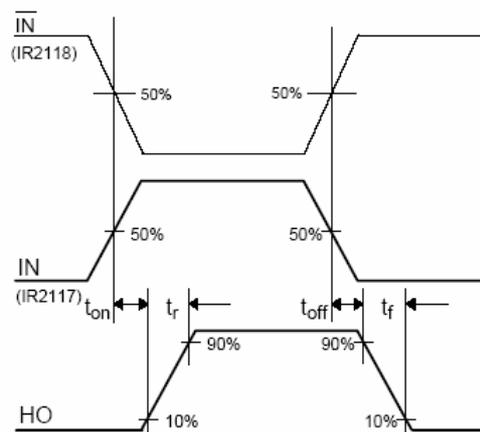


Figure 4. Switching Time Waveform Definition

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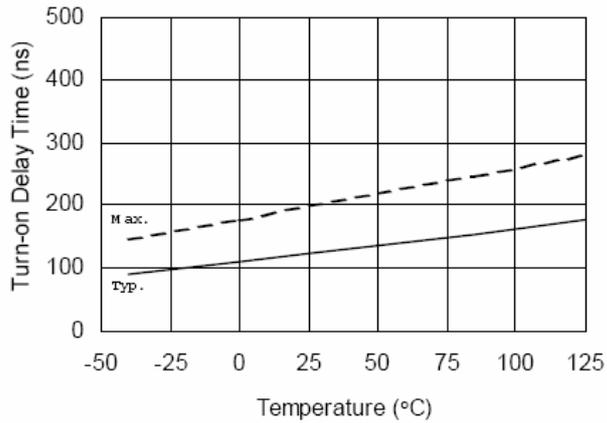


Figure 4A. Turn-On Time vs. Temperature

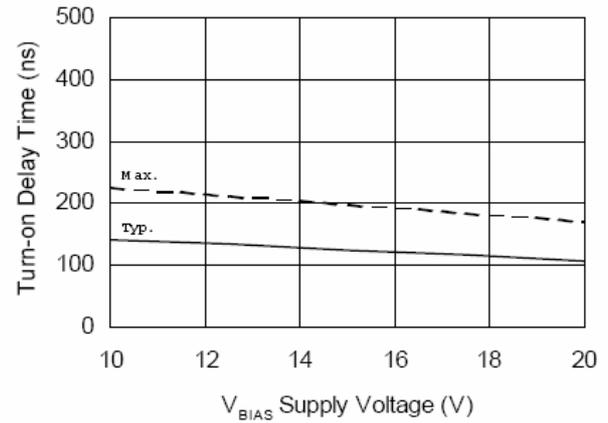


Figure 4B. Turn-On Time vs. Supply Voltage

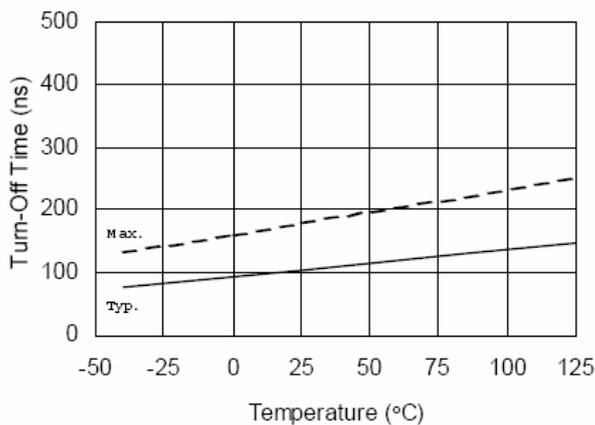


Figure 5A. Turn-Off Time vs. Temperature

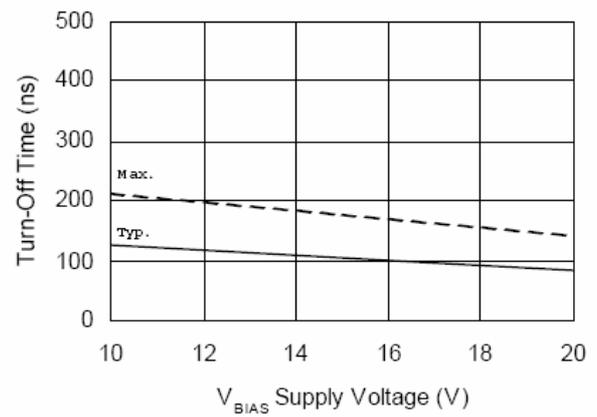
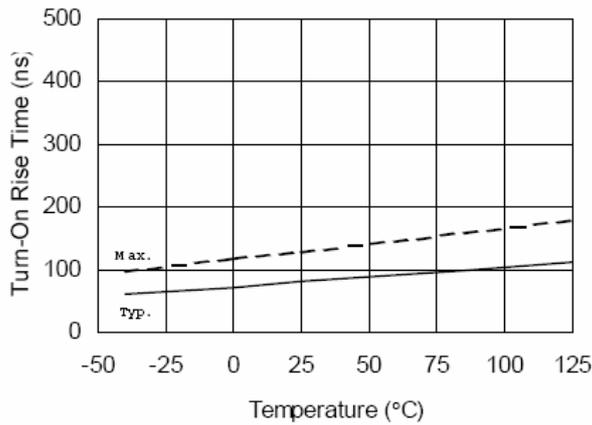


Figure 5B. Turn-Off Time vs. Supply Voltage

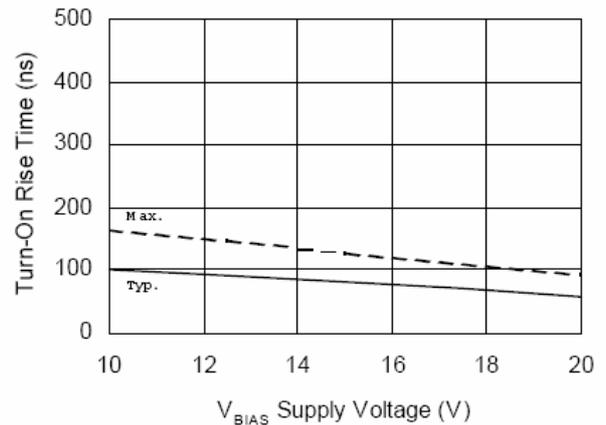
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	Diseño e implementación de un convertidor monofásico de cinco niveles con control basado en DSP	José Francisco Campos Bizcocho

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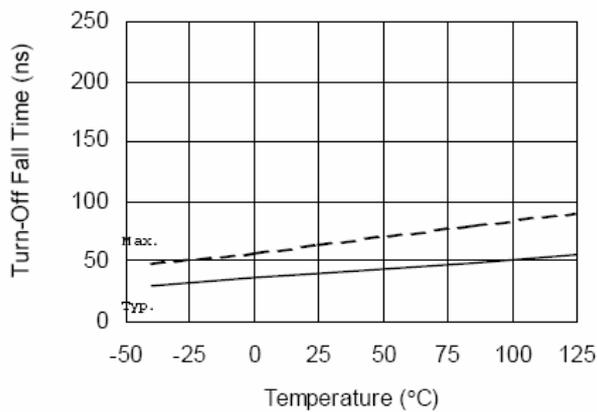
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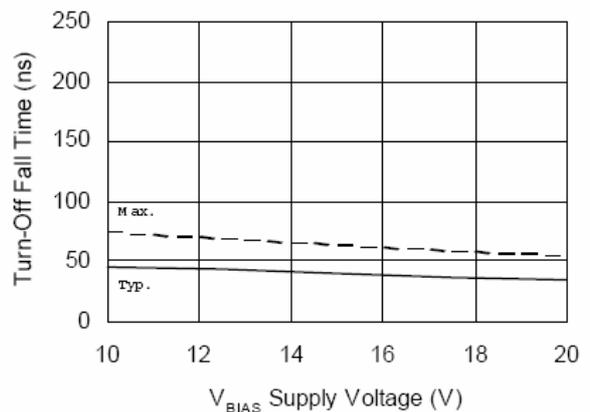
**Figure 6A. Turn-On Rise Time vs. Temperature**



**Figure 6B. Turn-On Rise Time vs. Supply Voltage**

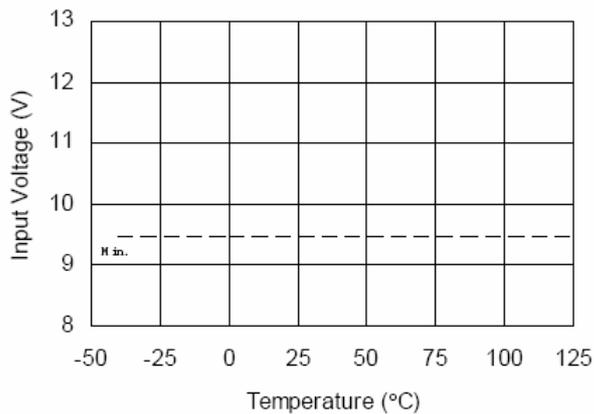


**Figure 7A. Turn-Off Fall Time vs. Temperature**

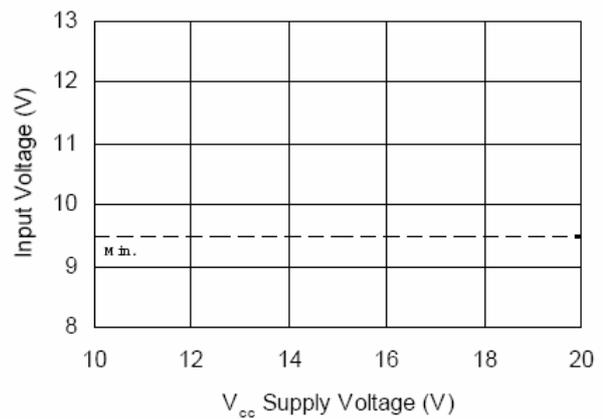


**Figure 7B. Turn-Off Fall Time vs. Supply Voltage**

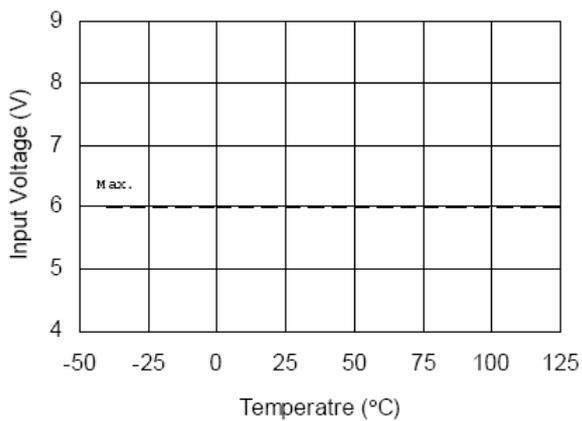
## IR2117(S)/IR2118(S) & (PbF)



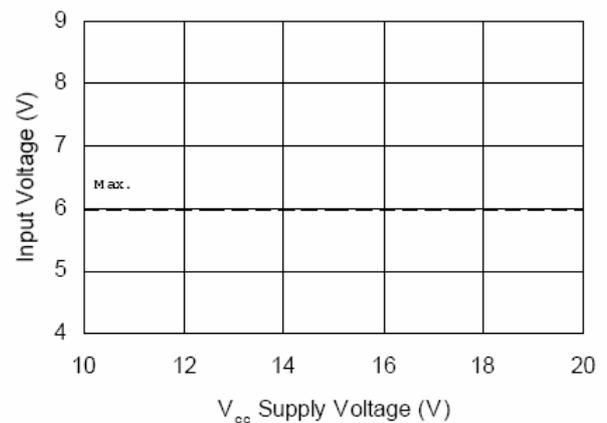
**Figure 8A. Logic "1" (IR2118 "0") Input Voltage vs. Temperature**



**Figure 8B. Logic "1" (IR2118 "0") Input Voltage vs. Supply Voltage**



**Figure 9A. Logic "0" (IR2118 "1") Input Voltage vs. Temperature**

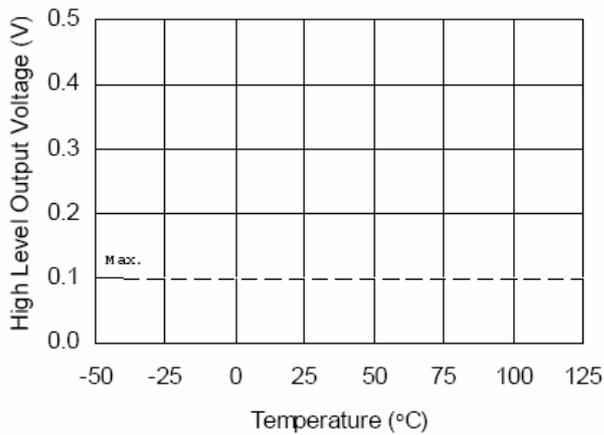


**Figure 9B. Logic "0" (IR2118 "1") Input Voltage vs. Supply Voltage**

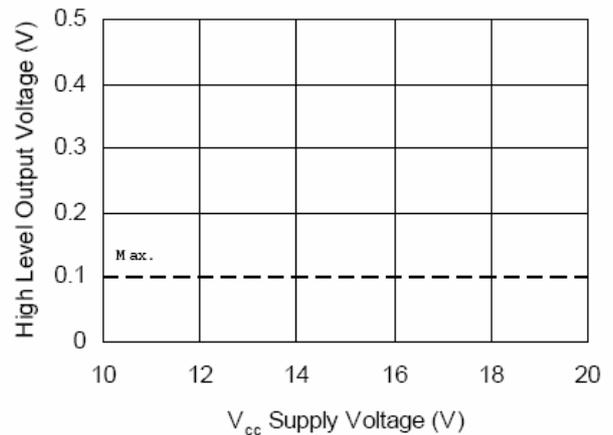
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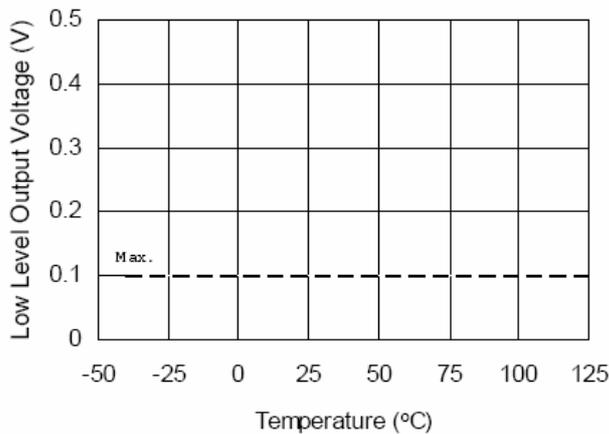
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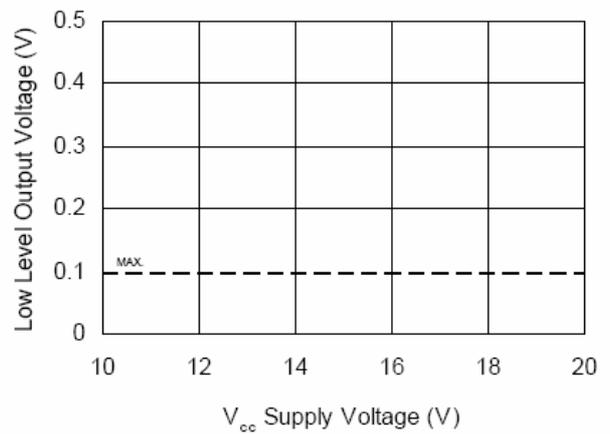
**Figure 10A. High Level Output vs. Temperature**



**Figure 10B. High Level Output vs. Supply Voltage**



**Figure 11A. Low Level Output vs. Temperature**



**Figure 11B. Low Level Output vs. Supply Voltage**

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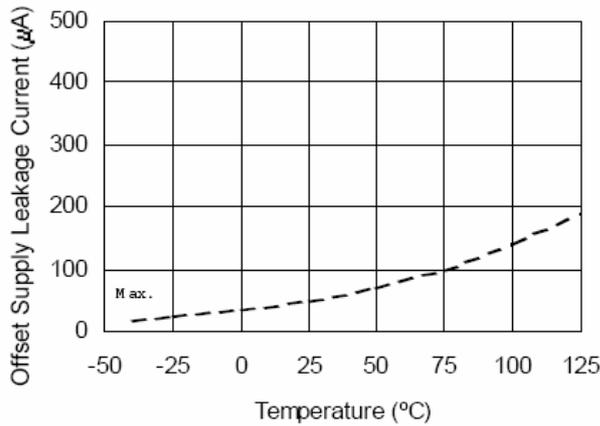


Figure 12A. Offset Supply Leakage Current vs. Temperature

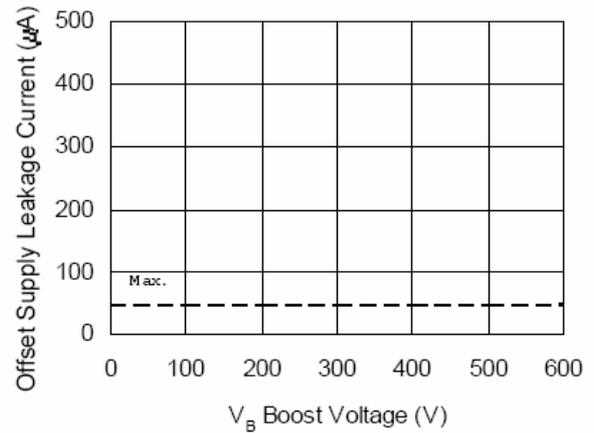


Figure 12B. Offset Supply Leakage Current vs. V<sub>B</sub> Boost Voltage

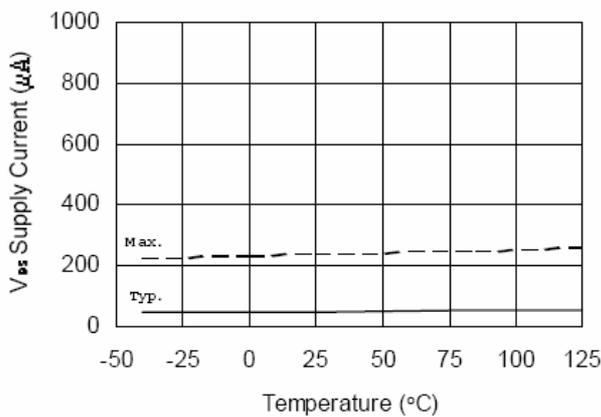


Figure 13A. V<sub>BS</sub> Supply Current vs. Temperature

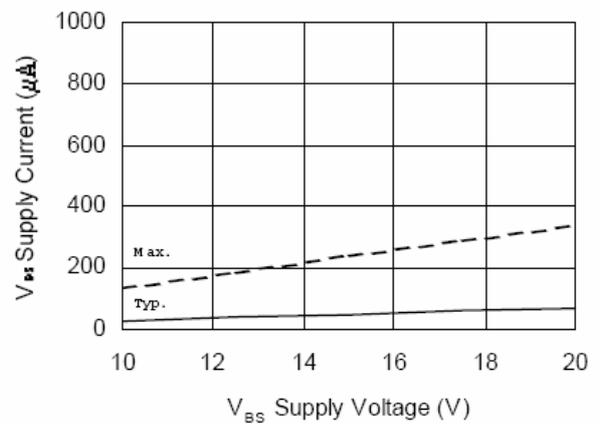


Figure 13B. V<sub>BS</sub> Supply Current vs. Supply Voltage

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	Diseño e implementación de un convertidor monofásico de cinco niveles con control basado en DSP	José Francisco Campos Bizcocho

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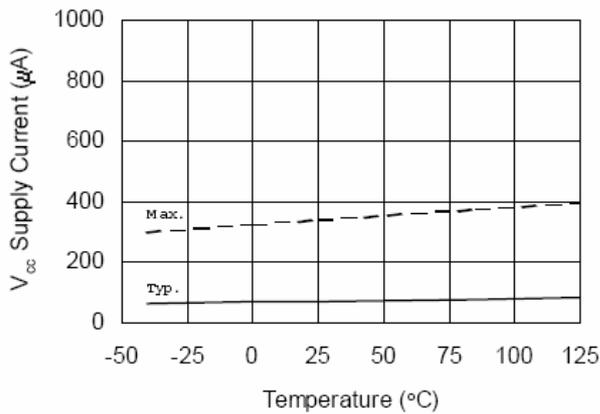


Figure 14A.  $V_{cc}$  Supply Current vs. Temperature

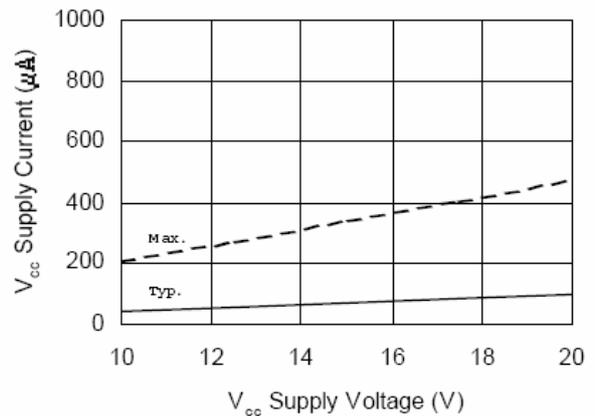


Figure 14B.  $V_{cc}$  Supply Current vs. Supply Voltage

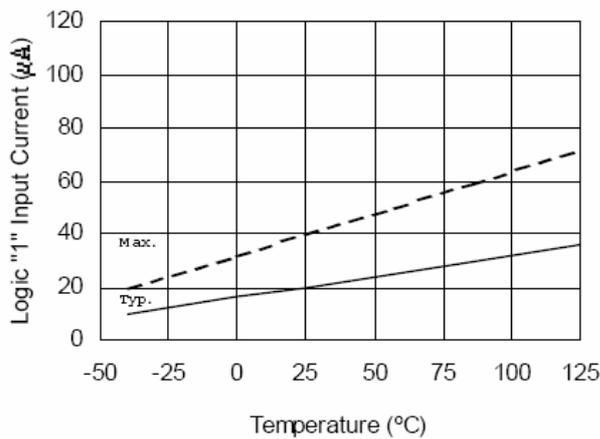


Figure 15A. Logic "1" (2118 "0") Input Current vs. Temperature

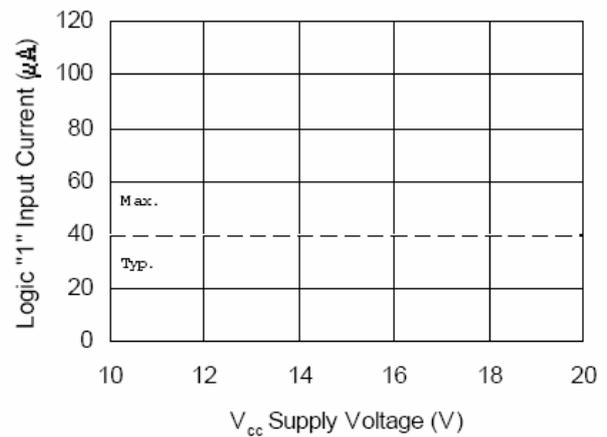


Figure 15B. Logic "1" (2118 "0") Input Current vs. Supply Voltage

	Proyecto Fin de Carrera	Alumno
	Diseño e implementación de un convertidor monofásico de cinco niveles con control basado en DSP	José Francisco Campos Bizcocho

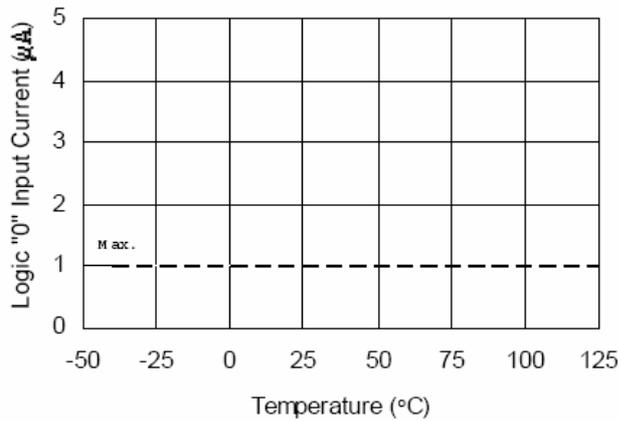


Figure 16A. Logic "0" (2118"1") Input Current vs. Temperature

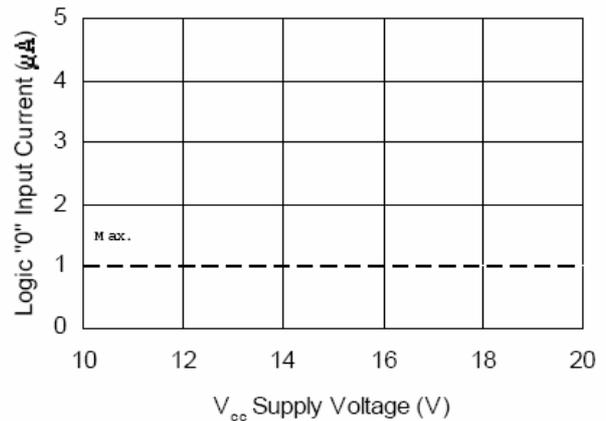


Figure 16B. Logic "0" (2118"1") Input Current vs. Supply Voltage

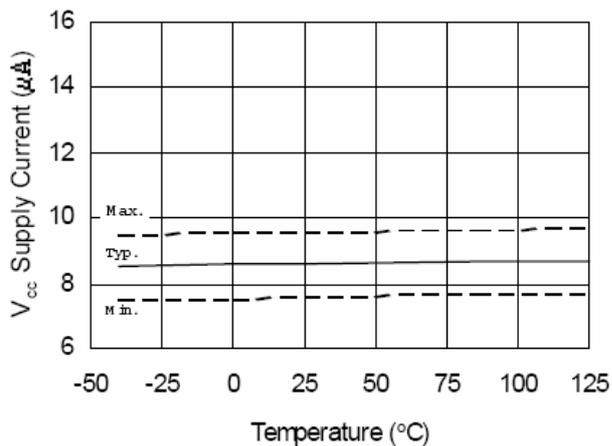


Figure 17A. V<sub>cc</sub> Undervoltage Threshold (+) vs. Temperature

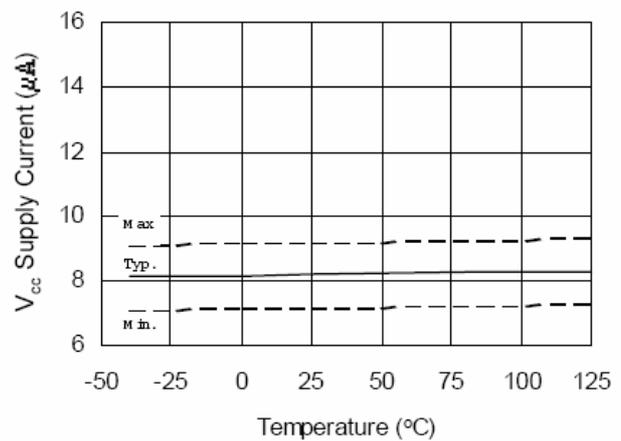


Figure 18A. V<sub>cc</sub> Undervoltage Threshold (-) vs. Temperature

	Proyecto Fin de Carrera	Alumno
	Diseño e implementación de un convertidor monofásico de cinco niveles con control basado en DSP	José Francisco Campos Bizcocho

## IR2117(S)/IR2118(S) & (PbF)

International  
**IR** Rectifier

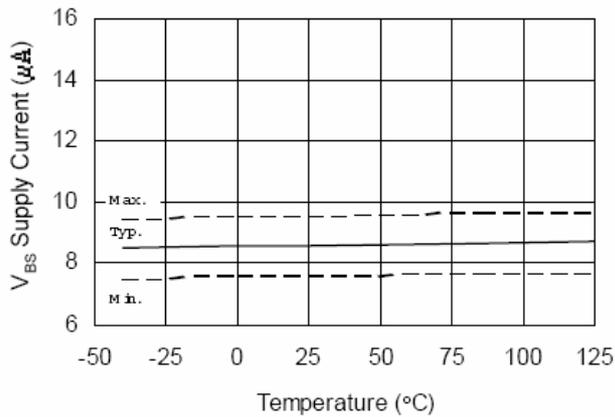


Figure 19A.  $V_{BS}$  Undervoltage Threshold (+) vs. Temperature

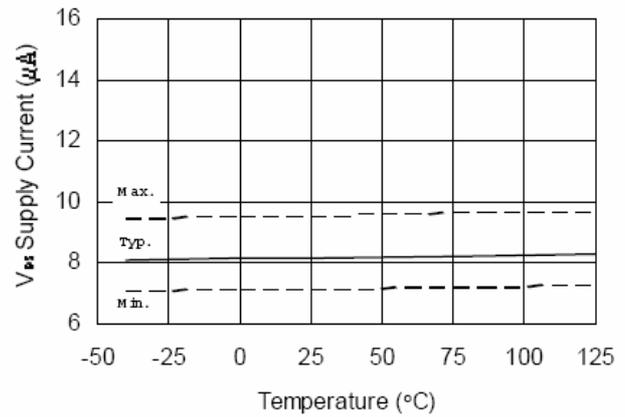


Figure 20A.  $V_{BS}$  Undervoltage Threshold (-) vs. Temperature

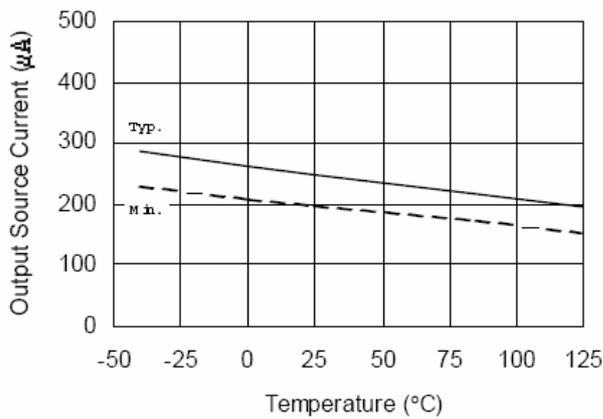


Figure 21A. Output Source Current vs. Temperature

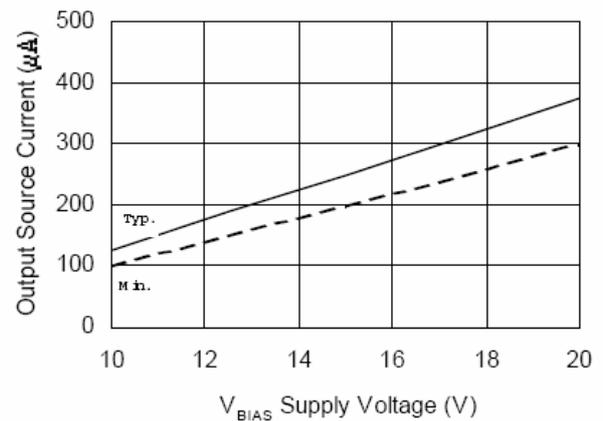


Figure 21B. Output Source Current vs. Supply Voltage

International  
 Rectifier

## IR2117(S)/IR2118(S) & (PbF)

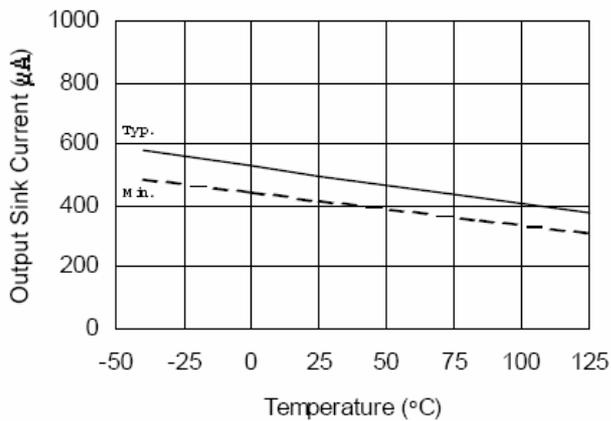


Figure 22A. Output Sink Current vs. Temperature

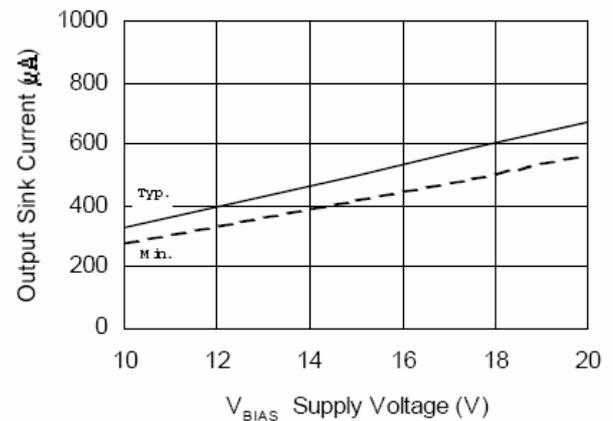


Figure 22B. Output Sink Current vs. Supply Voltage

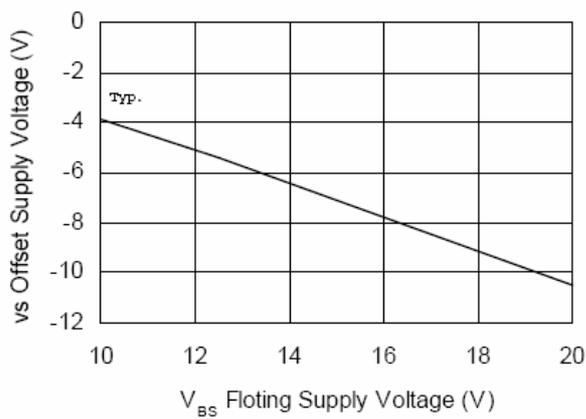
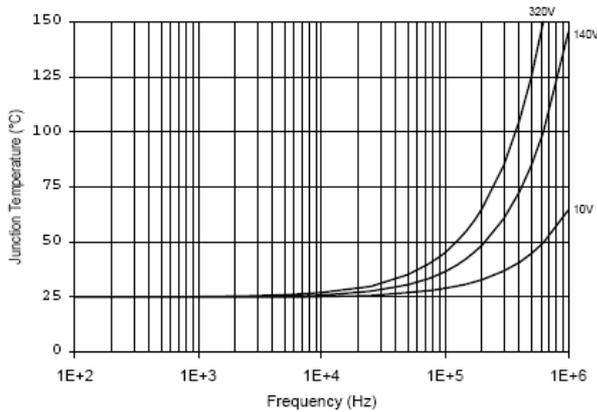


Figure 23B. Maximum VS Negative Offset vs. Supply Voltage

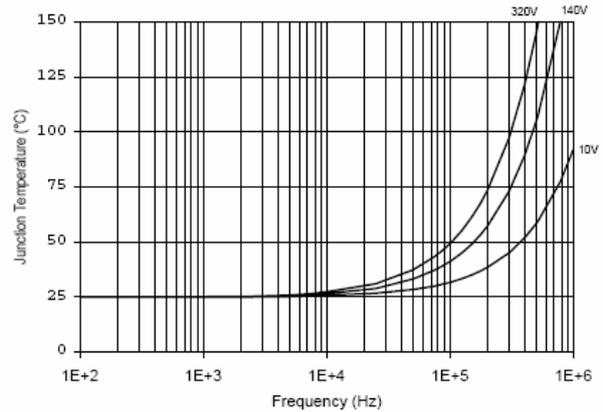
	Proyecto Fin de Carrera	Alumno
	Diseño e implementación de un convertidor monofásico de cinco niveles con control basado en DSP	José Francisco Campos Bizcocho

# IR2117(S)/IR2118(S) & (PbF)

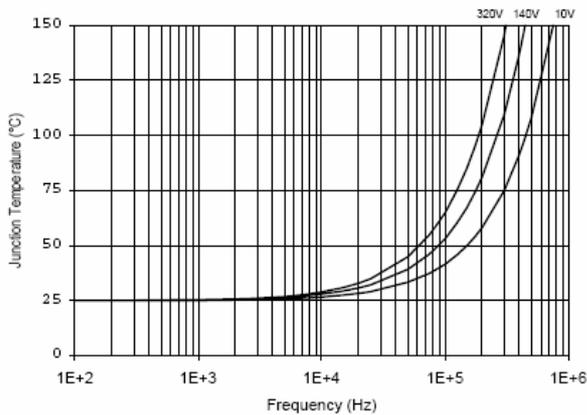
International  
**IR** Rectifier



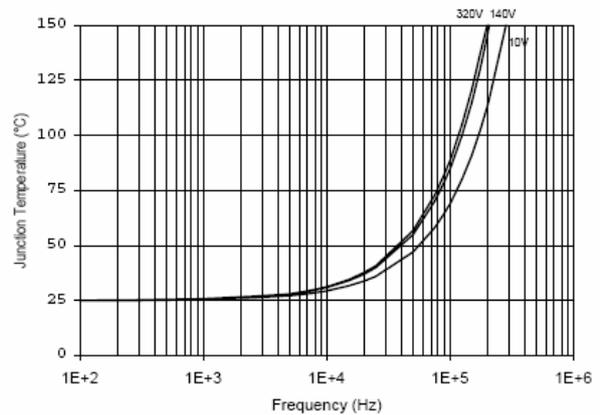
**Figure 24. IR2117/IR2118 T<sub>J</sub> vs. Frequency (IRFBC20)**  
R<sub>GATE</sub> = 33Ω, V<sub>CC</sub> = 15V



**Figure 25. IR2117/IR2118 T<sub>J</sub> vs. Frequency (IRFBC30)**  
R<sub>GATE</sub> = 22Ω, V<sub>CC</sub> = 15V

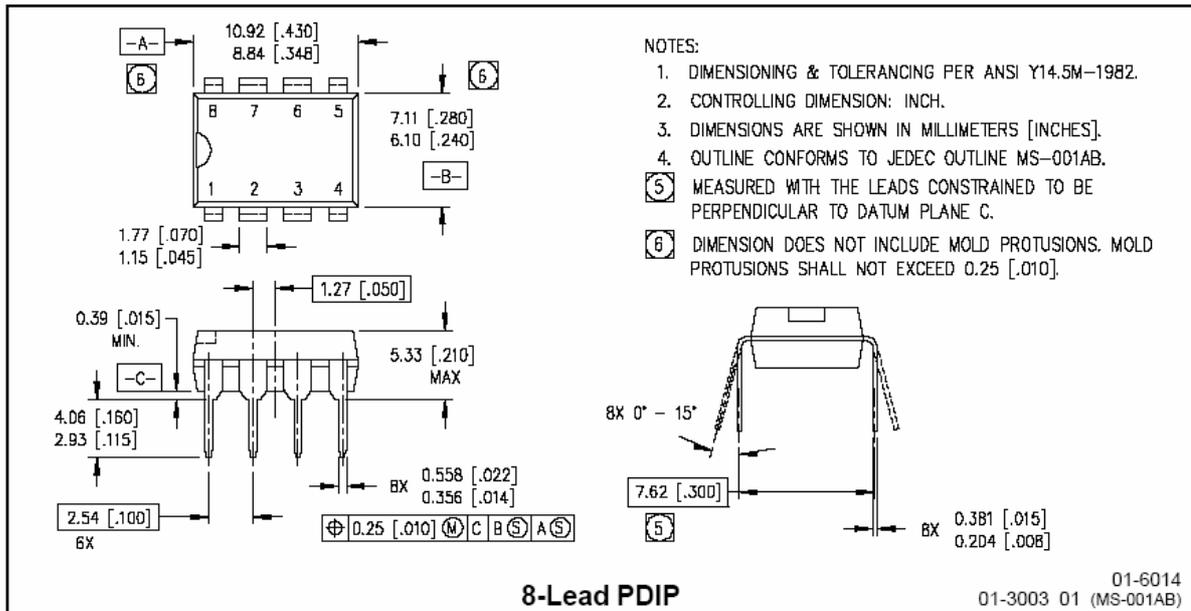


**Figure 26. IR2117/IR2118 T<sub>J</sub> vs. Frequency (IRFBC40)**  
R<sub>GATE</sub> = 15Ω, V<sub>CC</sub> = 15V

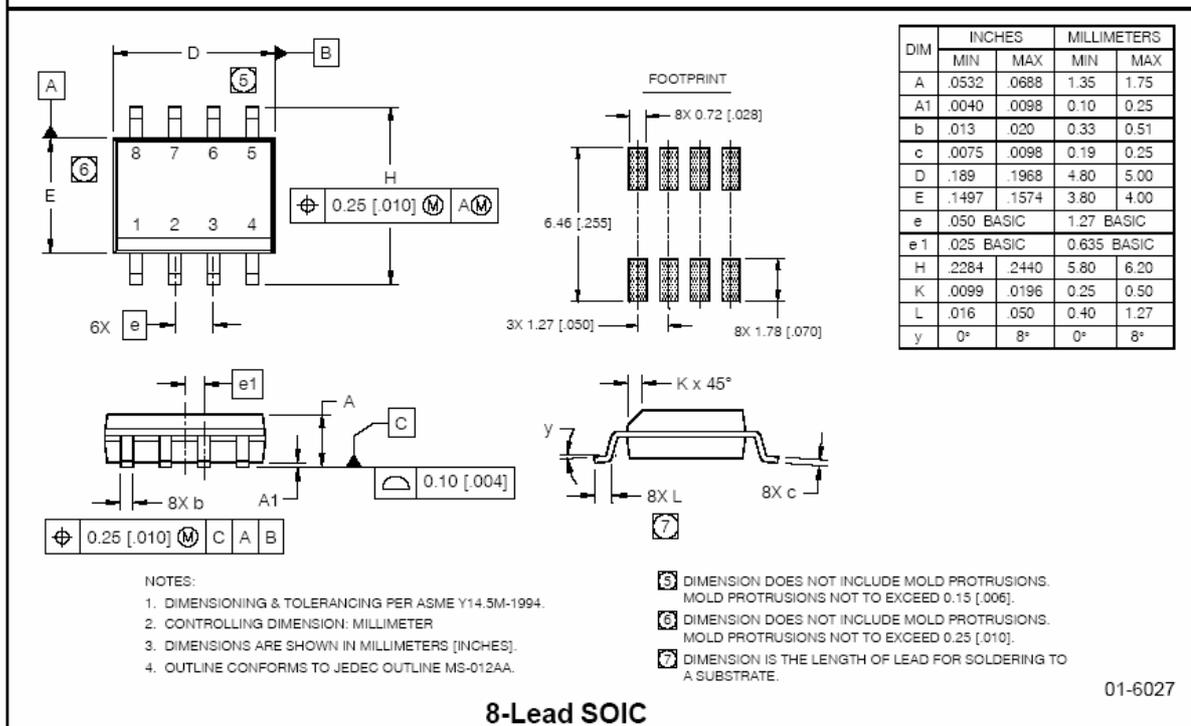


**Figure 27. IR2117/IR2118 T<sub>J</sub> vs. Frequency (IRFPE50)**  
R<sub>GATE</sub> = 10Ω, V<sub>CC</sub> = 15V

## Case outlines



**8-Lead PDIP**



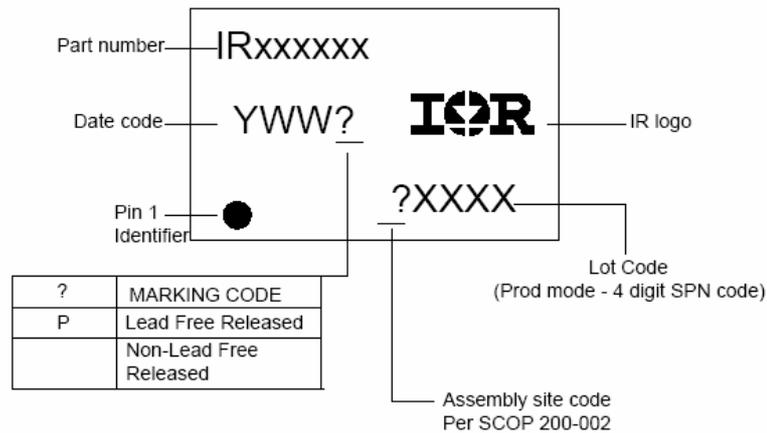
**8-Lead SOIC**

	Proyecto Fin de Carrera	Alumno
	Diseño e implementación de un convertidor monofásico de cinco niveles con control basado en DSP	José Francisco Campos Bizcocho

# IR2117(S)/IR2118(S) & (PbF)

International  
**IOR** Rectifier

## LEADFREE PART MARKING INFORMATION



## ORDER INFORMATION

### Basic Part (Non-Lead Free)

8-Lead PDIP IR2117 order IR2117  
 8-Lead PDIP IR2118 order IR2118  
 8-Lead SOIC IR2117S order IR2117S  
 8-Lead SOIC IR2118S order IR2118S

### Leadfree Part

8-Lead PDIP IR2117 order IR2117PbF  
 8-Lead PDIP IR2118 order IR2118PbF  
 8-Lead SOIC IR2117S order IR2117SPbF  
 8-Lead SOIC IR2118S order IR2118SPbF

International  
**IOR** Rectifier

IR WORLD HEADQUARTERS: 233 Kansas St., El Segundo, California 90245 Tel: (310) 252-7105

**This product has been qualified per industrial level**

*Data and specifications subject to change without notice. 4/2/2004*

## 4. H11L1.



### 6-Pin DIP Optoisolators Logic Output

The H11L1 and H11L2 have a gallium arsenide IRED optically coupled to a high-speed integrated detector with Schmitt trigger output. Designed for applications requiring electrical isolation, fast response time, noise immunity and digital logic compatibility.

- Guaranteed Switching Times —  $t_{on}, t_{off} < 4 \mu s$
- Built-In On/Off Threshold Hysteresis
- High Data Rate, 1 MHz Typical (NRZ)
- Wide Supply Voltage Capability
- Microprocessor Compatible Drive
- **To order devices that are tested and marked per VDE 0884 requirements, the suffix "V" must be included at end of part number. VDE 0884 is a test option.**

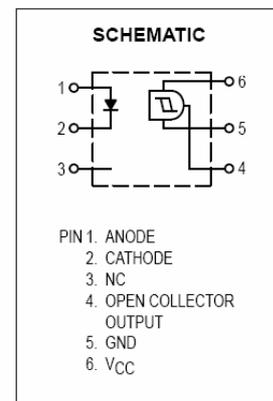
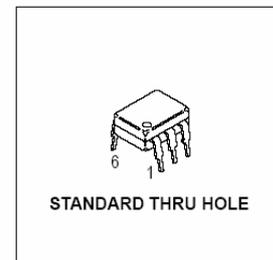
#### Applications

- Interfacing Computer Terminals to Peripheral Equipment
- Digital Control of Power Supplies
- Line Receiver — Eliminates Noise
- Digital Control of Motors and Other Servo Machine Applications
- Logic to Logic Isolator
- Logic Level Shifter — Couples TTL to CMOS

**MAXIMUM RATINGS** ( $T_A = 25^\circ C$  unless otherwise noted)

Rating	Symbol	Value	Unit
<b>INPUT LED</b>			
Reverse Voltage	$V_R$	6	Volts
Forward Current — Continuous	$I_F$	60	mA
— Peak		1.2	Amp
Pulse Width = 300 $\mu s$ , 2% Duty Cycle			
LED Power Dissipation @ $T_A = 25^\circ C$	$P_D$	120	mW
Derate above $25^\circ C$		1.41	mW/ $^\circ C$
<b>OUTPUT DETECTOR</b>			
Output Voltage Range	$V_O$	0–16	Volts
Supply Voltage Range	$V_{CC}$	3–16	Volts
Output Current	$I_O$	50	mA
Detector Power Dissipation @ $T_A = 25^\circ C$	$P_D$	150	mW
Derate above $25^\circ C$		1.76	mW/ $^\circ C$
<b>TOTAL DEVICE</b>			
Total Device Dissipation @ $T_A = 25^\circ C$	$P_D$	250	mW
Derate above $25^\circ C$		2.94	mW/ $^\circ C$
Maximum Operating Temperature	$T_A$	-40 to +85	$^\circ C$
Storage Temperature Rang	$T_{stg}$	-55 to +150	$^\circ C$
Soldering Temperature (10 s)	$T_L$	260	$^\circ C$
Isolation Surge Voltage (Pk ac Voltage, 60 Hz, 1 Second Duration)(1)	$V_{ISO}$	7500	Vac(pk)

1. Isolation surge voltage is an internal device dielectric breakdown rating.  
For this test, Pins 1 and 2 are common, and Pins 4, 5 and 6 are common.



**ELECTRICAL CHARACTERISTICS** ( $T_A = 25^\circ\text{C}$  unless otherwise noted)<sup>(1)</sup>

Characteristic	Symbol	Min	Typ <sup>(1)</sup>	Max	Unit	
<b>INPUT LED</b>						
Reverse Leakage Current ( $V_R = 3\text{ V}$ , $R_L = 1\text{ M}\Omega$ )	$I_R$	—	0.05	10	$\mu\text{A}$	
Forward Voltage ( $I_F = 10\text{ mA}$ ) ( $I_F = 0.3\text{ mA}$ )	$V_F$	— 0.75	1.2 0.95	1.5 —	Volts	
Capacitance ( $V_R = 0\text{ V}$ , $f = 1\text{ MHz}$ )	C	—	18	—	pF	
<b>OUTPUT DETECTOR</b>						
Operating Voltage	$V_{CC}$	3	—	15	Volts	
Supply Current ( $I_F = 0$ , $V_{CC} = 5\text{ V}$ )	$I_{CC(\text{off})}$	—	1	5	mA	
Output Current, High ( $I_F = 0$ , $V_{CC} = V_o = 15\text{ V}$ )	$I_{OH}$	—	—	100	$\mu\text{A}$	
<b>COUPLED</b>						
Supply Current ( $I_F = I_{F(\text{on})}$ , $V_{CC} = 5\text{ V}$ )	$I_{CC(\text{on})}$	—	1.6	5	mA	
Output Voltage, Low ( $R_L = 270\ \Omega$ , $V_{CC} = 5\text{ V}$ , $I_F = I_{F(\text{on})}$ )	$V_{OL}$	—	0.2	0.4	Volts	
Threshold Current, ON ( $R_L = 270\ \Omega$ , $V_{CC} = 5\text{ V}$ )	H11L1 H11L2 $I_{F(\text{on})}$	— —	1.2 —	1.6 10	mA	
Threshold Current, OFF ( $R_L = 270\ \Omega$ , $V_{CC} = 5\text{ V}$ )	H11L1 H11L2 $I_{F(\text{off})}$	0.3 0.3	0.75 —	— —	mA	
Hysteresis Ratio ( $R_L = 270\ \Omega$ , $V_{CC} = 5\text{ V}$ )	$\frac{I_{F(\text{off})}}{I_{F(\text{on})}}$	0.5	0.75	0.9		
Isolation Voltage <sup>(2)</sup> 60 Hz, AC Peak, 1 second, $T_A = 25^\circ\text{C}$	$V_{ISO}$	7500	—	—	Vac(pk)	
Turn-On Time	$R_L = 270\ \Omega^{(3)}$ $V_{CC} = 5\text{ V}$ , $I_F = I_{F(\text{on})}$ $T_A = 25^\circ\text{C}$	$t_{on}$	—	1.2	4	$\mu\text{s}$
Fall Time		$t_f$	—	0.1	—	
Turn-Off Time		$t_{off}$	—	1.2	4	
Rise Time		$t_r$	—	0.1	—	

1. Always design to the specified minimum/maximum electrical limits (where applicable).
2. For this test, IRED Pins 1 and 2 are common and Output Gate Pins 4, 5, 6 are common.
3.  $R_L$  value effect on switching time is negligible.

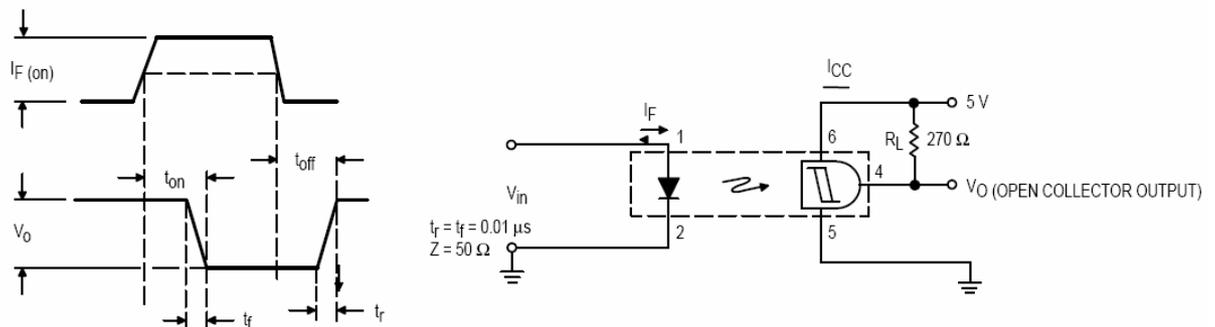


Figure 1. Switching Test Circuit

TYPICAL CHARACTERISTICS

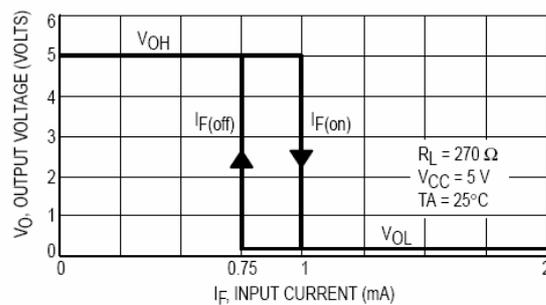


Figure 2. Transfer Characteristics for H11L1

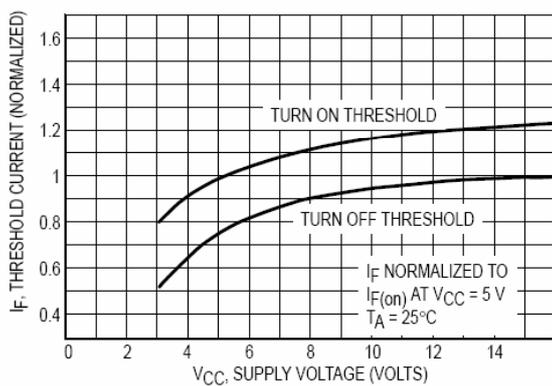


Figure 3. Threshold Current versus Supply Voltage

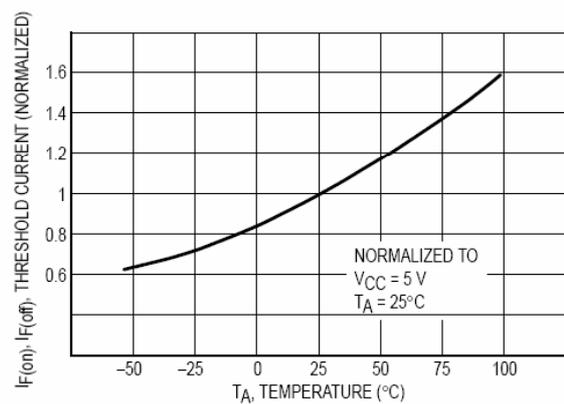


Figure 4. Threshold Current versus Temperature

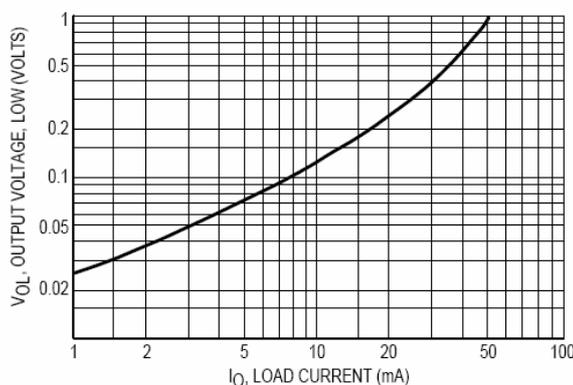


Figure 5. Output Voltage, Low versus Load Current

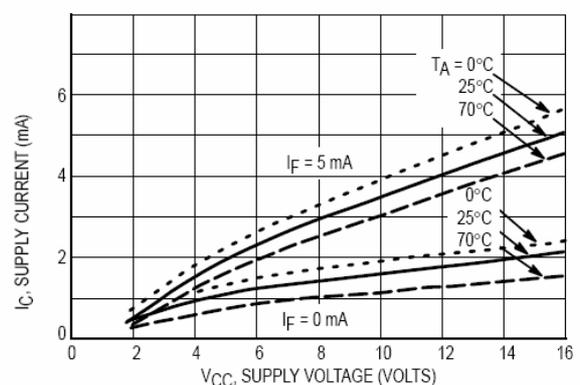
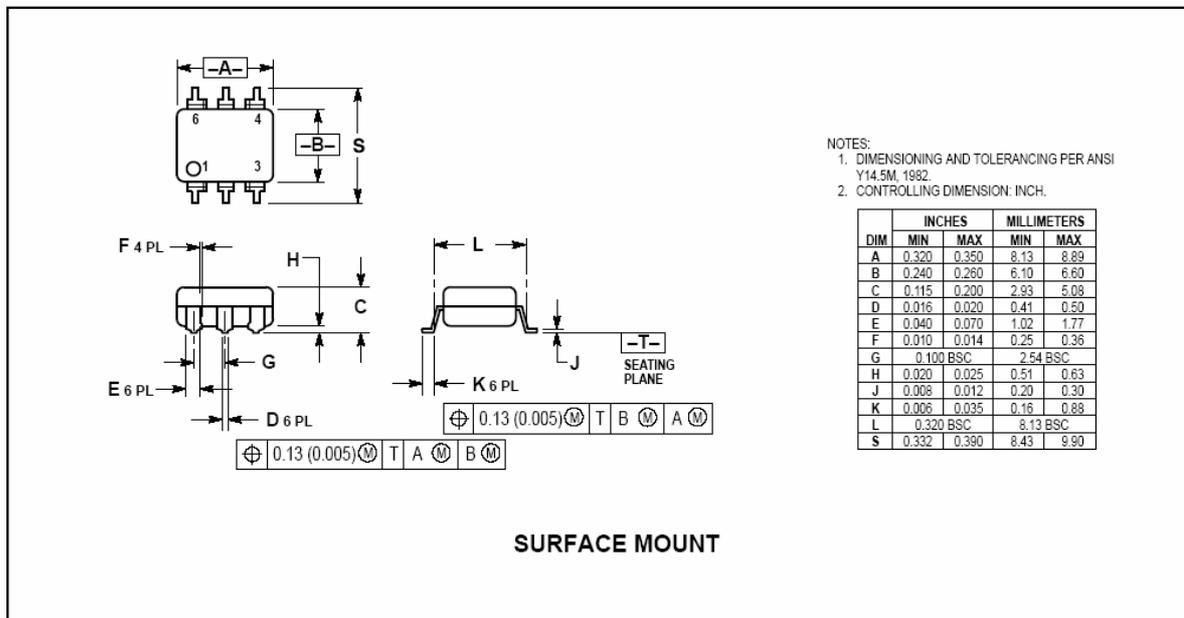
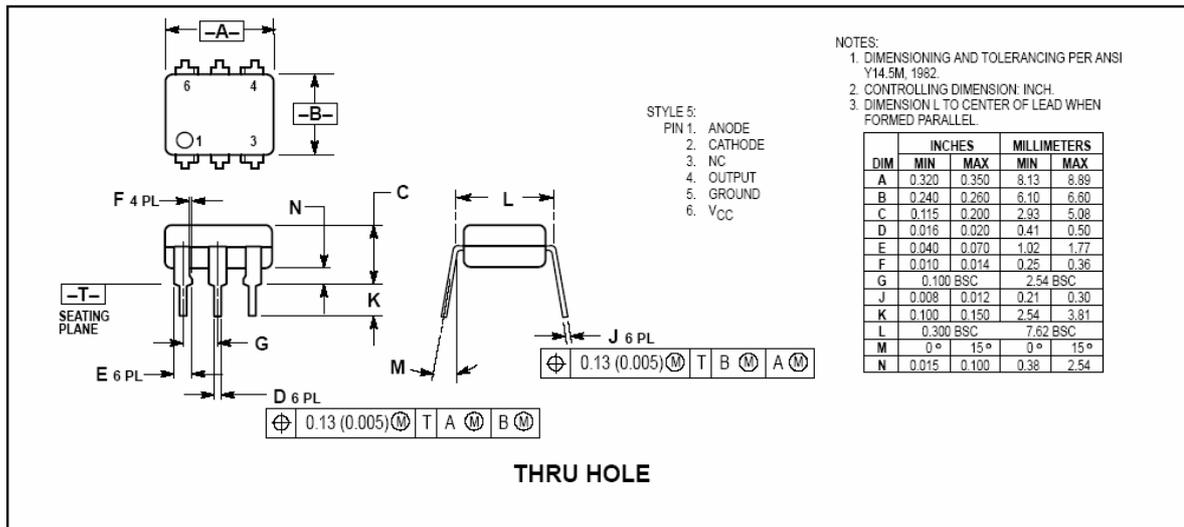


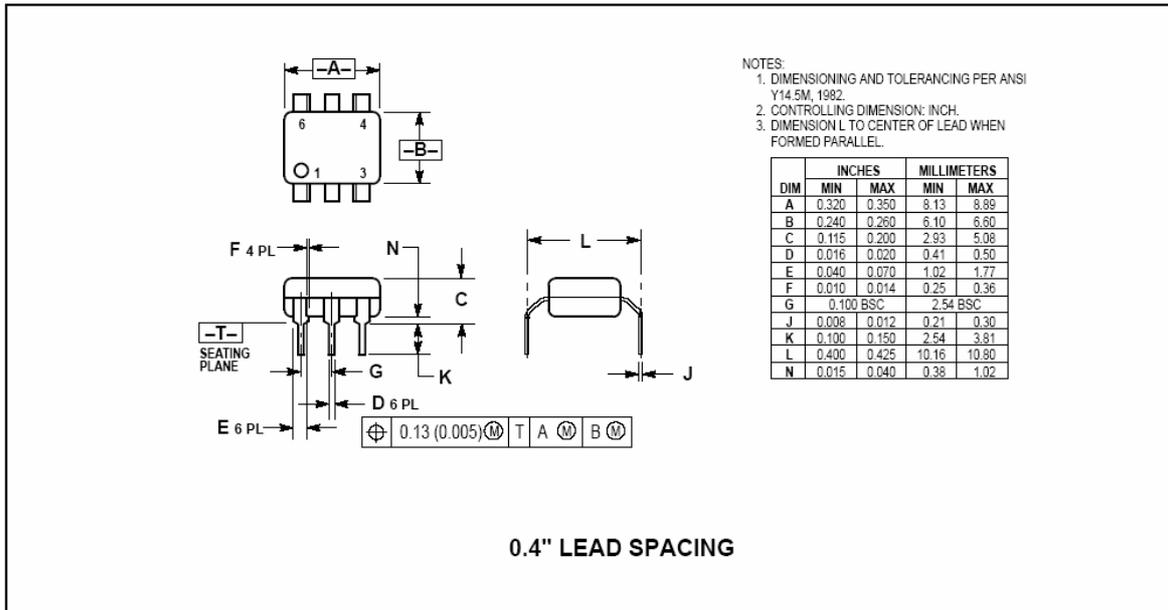
Figure 6. Supply Current versus Supply Voltage



# H11L1, H11L2

## PACKAGE DIMENSIONS





	Proyecto Fin de Carrera	Alumno
	Diseño e implementación de un convertidor monofásico de cinco niveles con control basado en DSP	José Francisco Campos Bizcocho



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2. A critical component in any component of a life support device or system whose failure to perform can be reasonably expected to cause the failure of the life support device or system, or to affect its safety or effectiveness.

## 5. LV 25-P.

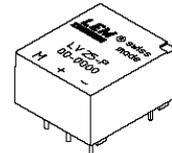


### Voltage Transducer LV 25-P

For the electronic measurement of voltages : DC, AC, pulsed..., with a galvanic isolation between the primary circuit (high voltage) and the secondary circuit (electronic circuit).

$$I_{PN} = 10 \text{ mA}$$

$$V_{PN} = 10 \dots 500 \text{ V}$$



#### Electrical data

$I_{PN}$	Primary nominal r.m.s. current	10	mA			
$I_P$	Primary current, measuring range	0 .. ± 14	mA			
$R_M$	Measuring resistance	with ± 12 V	@ ± 10 mA <sub>max</sub>	$R_{Mmin}$	$R_{Mmax}$	
			@ ± 14 mA <sub>max</sub>	30	190	Ω
	with ± 15 V	@ ± 10 mA <sub>max</sub>	30	100	Ω	
		@ ± 14 mA <sub>max</sub>	100	350	Ω	
$I_{SN}$	Secondary nominal r.m.s. current	25	mA			
$K_N$	Conversion ratio	2500 : 1000				
$V_C$	Supply voltage (± 5 %)	± 12 .. 15	V			
$I_C$	Current consumption	10 (@±15V)+ $I_S$	mA			
$V_d$	R.m.s. voltage for AC isolation test <sup>1)</sup> , 50 Hz, 1 mn	2.5	kV			

#### Features

- Closed loop (compensated) voltage transducer using the Hall effect
- Insulated plastic case recognized according to UL 94-V0.

#### Principle of use

- For voltage measurements, a current proportional to the measured voltage must be passed through an external resistor  $R_x$  which is selected by the user and installed in series with the primary circuit of the transducer.

#### Accuracy - Dynamic performance data

$X_G$	Overall Accuracy @ $I_{PN}, T_A = 25^\circ\text{C}$	@ ± 12 .. 15 V	± 0.9	%	
		@ ± 15 V (± 5 %)	± 0.8	%	
$\mathcal{E}_L$	Linearity		< 0.2	%	
$I_O$	Offset current @ $I_P = 0, T_A = 25^\circ\text{C}$	Typ	± 0.15	mA	
		Max	± 0.15	mA	
$I_{OT}$	Thermal drift of $I_O$	0°C .. + 25°C	± 0.06	± 0.25	mA
		+ 25°C .. + 70°C	± 0.10	± 0.35	mA
$t_r$	Response time <sup>2)</sup> @ 90 % of $V_{Pmax}$		40	μs	

#### Advantages

- Excellent accuracy
- Very good linearity
- Low thermal drift
- Low response time
- High bandwidth
- High immunity to external interference
- Low disturbance in common mode.

#### General data

$T_A$	Ambient operating temperature	0 .. + 70	°C
$T_S$	Ambient storage temperature	- 25 .. + 85	°C
$R_p$	Primary coil resistance @ $T_A = 70^\circ\text{C}$	250	Ω
$R_s$	Secondary coil resistance @ $T_A = 70^\circ\text{C}$	110	Ω
$m$	Mass	22	g
	Standards <sup>3)</sup>	EN 50178	

#### Applications

- AC variable speed drives and servo motor drives
- Static converters for DC motor drives
- Battery supplied applications
- Uninterruptible Power Supplies (UPS)
- Power supplies for welding applications.

Notes : <sup>1)</sup> Between primary and secondary

<sup>2)</sup>  $R_1 = 25 \text{ k}\Omega$  (L/R constant, produced by the resistance and inductance of the primary circuit)

<sup>3)</sup> A list of corresponding tests is available

981125/14

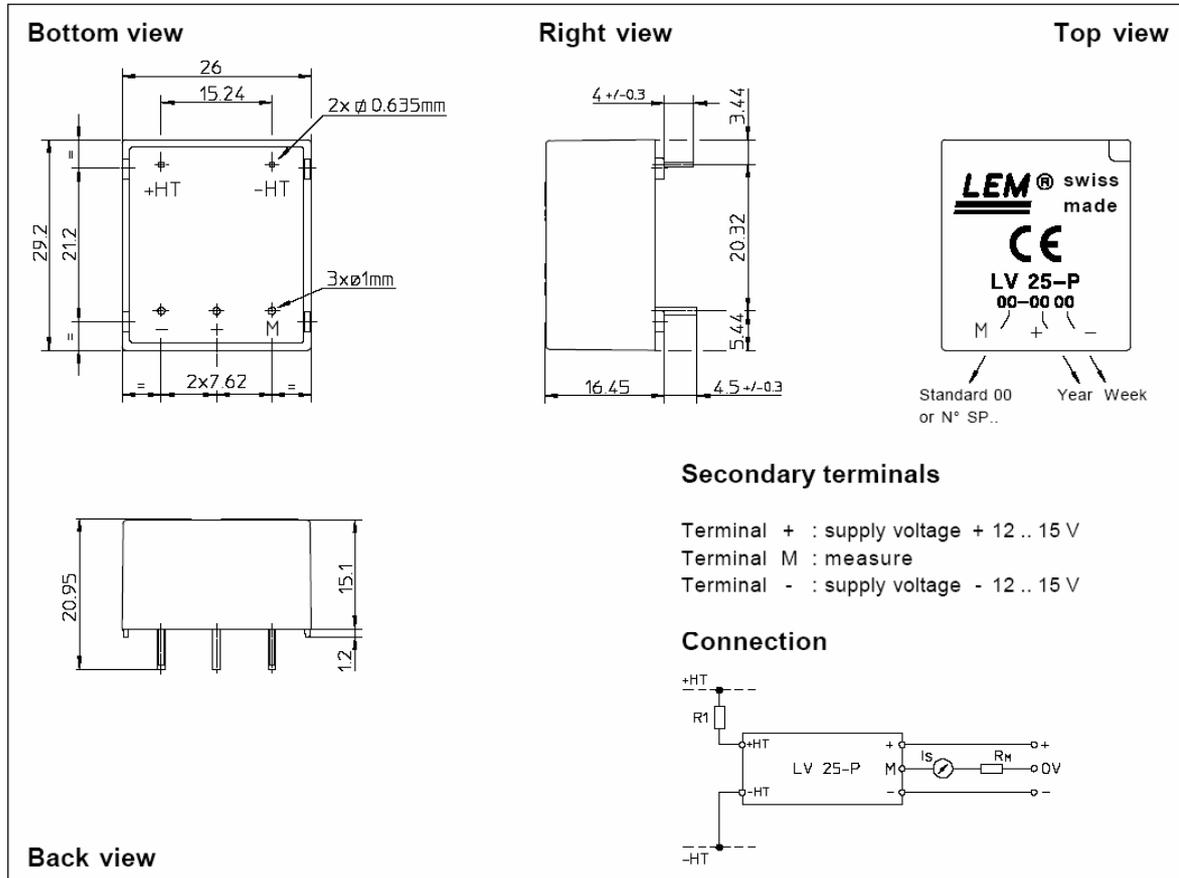
LEM Components

www.lem.com

Tope Co., Ltd. Tel: (02) 8228-0658 Fax: (02) 8228-0659 http://www.sensor.com.tw e-mail: tope@ms1.hinet.net



### Dimensions LV 25-P (in mm. 1 mm = 0.0394 inch)



### Mechanical characteristics

- General tolerance  $\pm 0.2 \text{ mm}$
- Fastening & connection of primary 2 pins  
0.635 x 0.635 mm
- Fastening & connection of secondary 3 pins  $\varnothing 1 \text{ mm}$
- Recommended PCB hole 1.2 mm

### Remarks

- $I_s$  is positive when  $V_p$  is applied on terminal +HT.
- This is a standard model. For different versions (supply voltages, turns ratios, unidirectional measurements...), please contact us.

### Instructions for use of the voltage transducer model LV 25-P

Primary resistor  $R_1$  : the transducer's optimum accuracy is obtained at the nominal primary current. As far as possible,  $R_1$  should be calculated so that the nominal voltage to be measured corresponds to a primary current of 10 mA.

Example: Voltage to be measured  $V_{PN} = 250 \text{ V}$

- a)  $R_1 = 25 \text{ k}\Omega / 2.5 \text{ W}$ ,  $I_p = 10 \text{ mA}$  Accuracy =  $\pm 0.8 \%$  of  $V_{PN}$  (@  $T_A = +25^\circ\text{C}$ )  
b)  $R_1 = 50 \text{ k}\Omega / 1.25 \text{ W}$ ,  $I_p = 5 \text{ mA}$  Accuracy =  $\pm 1.6 \%$  of  $V_{PN}$  (@  $T_A = +25^\circ\text{C}$ )

Operating range (recommended) : taking into account the resistance of the primary windings (which must remain low compared to  $R_1$ , in order to keep thermal deviation as low as possible) and the isolation, this transducer is suitable for measuring nominal voltages from 10 to 500 V.

LEM reserves the right to carry out modifications on its transducers, in order to improve them, without previous notice.

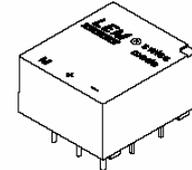
## 6. LA 25-NP.



### Current Transducer LA 25-NP

$I_{PN} = 5-6-8-12-25 \text{ A}$

For the electronic measurement of currents : DC, AC, pulsed, mixed, with a galvanic isolation between the primary circuit (high power) and the secondary circuit (electronic circuit).



#### Electrical data

$I_{PN}$	Primary nominal r.m.s. current	25	At
$I_P$	Primary current, measuring range	0 .. $\pm 36$	At
$R_M$	Measuring resistance @	$T_A = 70^\circ\text{C}$	$T_A = 85^\circ\text{C}$
		$R_{Mmin} \quad R_{Mmax}$	$R_{Mmin} \quad R_{Mmax}$
	with $\pm 15 \text{ V}$	@ $\pm 25 \text{ At}_{max}$	100 320 100 315 $\Omega$
		@ $\pm 36 \text{ At}_{max}$	100 190 100 185 $\Omega$
$I_{SN}$	Secondary nominal r.m.s. current	25	mA
$K_N$	Conversion ratio	1-2-3-4-5	: 1000
$V_C$	Supply voltage ( $\pm 5 \%$ )	$\pm 15$	V
$I_C$	Current consumption	$10 + I_S$	mA
$V_d$	R.m.s. voltage for AC isolation test, 50 Hz, 1 mn	2.5	kV
$V_b$	R.m.s. rated voltage <sup>1)</sup> , safe separation basic isolation	600	V
		1700	V

#### Features

- Closed loop (compensated) multi-range current transducer using the Hall effect
- Insulated plastic case recognized according to UL 94-V0.

#### Advantages

- Excellent accuracy
- Very good linearity
- Low temperature drift
- Optimized response time
- Wide frequency bandwidth
- No insertion losses
- High immunity to external interference
- Current overload capability.

#### Applications

- AC variable speed drives and servo motor drives
- Static converters for DC motor drives
- Battery supplied applications
- Uninterruptible Power Supplies (UPS)
- Switched Mode Power Supplies (SMPS)
- Power supplies for welding applications.

#### Accuracy - Dynamic performance data

$X$	Typical accuracy @ $I_{PN}, T_A = 25^\circ\text{C}$	$\pm 0.5$	%
$\epsilon_L$	Linearity error	< 0.2	%
$I_O$	Offset current <sup>2)</sup> @ $I_P = 0, T_A = 25^\circ\text{C}$	Typ	Max
		$\pm 0.05$	$\pm 0.15$ mA
$I_{OM}$	Residual current <sup>3)</sup> @ $I_P = 0$ , after an overload of $3 \times I_{PN}$	$\pm 0.05$	$\pm 0.15$ mA
$I_{OT}$	Thermal drift of $I_O$	0°C .. + 25°C	$\pm 0.06$ $\pm 0.25$ mA
		+ 25°C .. + 70°C	$\pm 0.10$ $\pm 0.35$ mA
		- 25°C .. + 85°C	$\pm 0.5$ mA
		- 40°C .. + 85°C	$\pm 1.2$ mA
$t_r$	Response time <sup>4)</sup> @ 90 % of $I_{PN}$	< 1	$\mu\text{s}$
$di/dt$	$di/dt$ accurately followed	> 50	A/ $\mu\text{s}$
$f$	Frequency bandwidth (- 1 dB)	DC .. 150	kHz

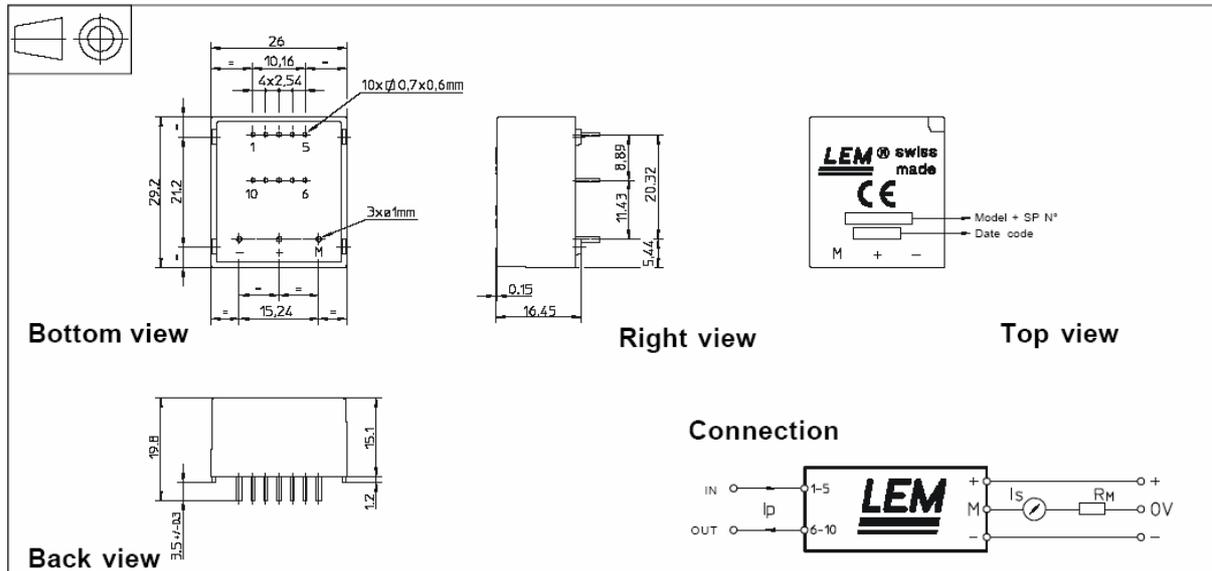
#### General data

$T_A$	Ambient operating temperature	- 40 .. + 85	$^\circ\text{C}$
$T_S$	Ambient storage temperature	- 45 .. + 90	$^\circ\text{C}$
$R_P$	Primary resistance per turn @ $T_A = 25^\circ\text{C}$	< 1.25	m $\Omega$
$R_S$	Secondary coil resistance @ $T_A = 70^\circ\text{C}$	110	$\Omega$
		@ $T_A = 85^\circ\text{C}$	115 $\Omega$
$R_{IS}$	Isolation resistance @ 500 V, $T_A = 25^\circ\text{C}$	> 1500	M $\Omega$
$m$	Mass	22	g
	Standards	EN 50178 : 1997	

- Notes : <sup>1)</sup> Pollution class 2  
<sup>2)</sup> Measurement carried out after 15 mn functioning  
<sup>3)</sup> The result of the coercive field of the magnetic circuit  
<sup>4)</sup> With a  $di/dt$  of 100 A/ $\mu\text{s}$ .



### Dimensions LA 25-NP (in mm. 1 mm = 0.0394 inch)



Number of primary turns	Primary current		Nominal output current $I_{SN}$ [mA]	Turns ratio $K_N$	Primary resistance $R_P$ [m $\Omega$ ]	Primary inductance $L_P$ [ $\mu$ H]	Recommended connections
	nominal $I_{PN}$ [A]	maximum $I_P$ [A]					
1	25	36	25	1/1000	0.3	0.023	
2	12	18	24	2/1000	1.1	0.09	
3	8	12	24	3/1000	2.5	0.21	
4	6	9	24	4/1000	4.4	0.37	
5	5	7	25	5/1000	6.3	0.58	

### Mechanical characteristics

- General tolerance  $\pm 0.2$  mm
- Fastening & connection of primary 10 pins 0.7 x 0.6 mm
- Fastening & connection of secondary 3 pins  $\varnothing 1$  mm
- Recommended PCB hole 1.2 mm

### Remarks

- $I_s$  is positive when  $I_p$  flows from terminals 1, 2, 3, 4, 5 to terminals 10, 9, 8, 7, 6
- This is a standard model. For different versions (supply voltages, turns ratios, unidirectional measurements...), please contact us.

	Proyecto Fin de Carrera	Alumno
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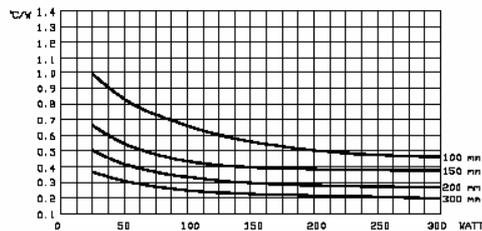
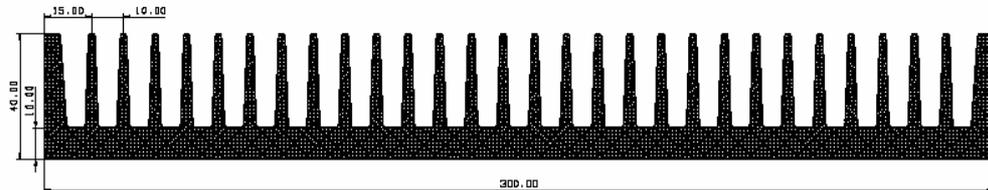
## 7. RG14092.



### RG14092

Radiador especialmente diseñado para el montaje de módulos de potencia estándar y componentes diversos.

**Peso: 15.73 Kgrs./metro.**  
**Perímetro: 2310 mm.**  
**Sección: 5823 mm<sup>2</sup>.**



ANOTACIONES DE USUARIO, EQUIVALENCIAS.			
FABRICANTE	REFERENCIA	FABRICANTE	REFERENCIA

RG14092

DISIPADORES/03/01

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## RG14092/ST

Radiador RG14092 con mecanizados y acabados estándar para su aplicación directa. Las dimensiones (longitud L) normalizadas son de 100, 150, 200 y 300 mm.

Las opciones de suministro son: N (anodizado negro), R (superficie fresada).

Resistencia térmica (valores prácticos)

RG14092/100N : 0.50 °C/W.	RG14092/100 : 0.165 °C/W.	(Ventilación forzada con una velocidad de 6 m/s.)
RG14092/150N : 0.38 °C/W.	RG14092/150 : 0.126 °C/W.	
RG14092/200N : 0.27 °C/W.	RG14092/200 : 0.09 °C/W.	
RG14092/300N : 0.21 °C/W.	RG14092/300 : 0.07 °C/W.	

Relación de referencias estándar:

RG14092/100, RG14092/100R, RG14092/100N, RG14092/100NR.  
 RG14092/150, RG14092/150R, RG14092/150N, RG14092/150NR.  
 RG14092/200, RG14092/200R, RG14092/200N, RG14092/200NR.  
 RG14092/300, RG14092/300R, RG14092/300N, RG14092/300NR.